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You can also visit article pages from back issues at www.nutsvolts.com. Just select the Back Issues tab from the Magazine drop down menu, click the Table of Contents link, and the article name.

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Low-cost Industrial Serial to Ethernet Solutions

- Instantly network-enable any serial device
- No programming is required for serial to Ethernet application
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Electronics Enthusiast - Is It Time To Redefine The Term?

I had lunch with a fellow electronics enthusiast on his return from a trip with his daughter in search of a college of engineering. He was blown away by the level of electronic automation in many of the campuses they visited. Not only was there ubiquitous Wi-Fi, but at several colleges, smartphone-toting students could monitor the status of the washing machines and dryers in each of the dorms. Tablet-sized LCD panels affixed to classroom doors announced activities in that room for the day, and numerous PC kiosks enabled students to check their email or class schedule. My friend’s daughter wasn’t as impressed with the modern electronic accoutrements, but instead expected what they found and more. Perhaps this explained my friend’s frustration with trying to teach his daughter the “old fashioned” electronics of resistors, transistors, and other discrete components. With sights set on creating a startup company using mobile computing, she couldn’t see the relevance learning to flip a few LEDs on and off with a complex circuit or of building a power supply from scratch when a single chip would do the job.

My automatic response was that learning the fundamentals is, well, fundamental. If you’re going to design or build electronics, you have to have a handle on the fundamentals. We decided that sooner or later, she would come around and delve into the basic physics and discrete components the way we did when we discovered electronics. However, I began to have second thoughts soon after the meeting. Is it really necessary to deal with capacitors, resistors, and discrete components to understand and — more importantly — leverage modern electronics? Is it a waste of time for someone planning to create a business based on new devices to play with and learn to use “outdated” components? Is it better to move directly to, say, microcontrollers and programmable field gate arrays (FPGAs)?

As an educator, I can say that knowledge of the fundamentals is important, but you have to be selective in the fundamentals you decide to focus on. I remember in college as an engineering major that knowledge of Quantum Mechanics was deemed critical to understanding how semiconductors work. So, I suffered through the course. To this day, I can’t say that I’ve applied any of what I might have learned to semiconductor design. On the other hand, if I had chosen to work for Intel or some other chip manufacturer, perhaps Quantum Mechanics would have served me well.

You no doubt have your own reasons for reading the articles in Nuts & Volts. You may be new to electronics, and every bit of information is new and exciting. Perhaps you’re retired and like the familiarity of the ‘old fashioned’ circuitry. Maybe you’re studying engineering and you’re using the articles on microcontrollers as supplemental material in your studies. Perhaps you simply like the satisfaction of building circuits with your hands. Many of your fellow readers are specialists in other fields and have an idea of automating some aspect of their work and look to this magazine to provide a soft introduction to the world of applied electronics.

After you read through this issue, I’d like you to email me with your thoughts on the overall focus of the magazine. Does it define — in your view — the world of the electronics enthusiast? As the field of electronics evolves, do you want us to cover the bleeding edge, or are you more comfortable working a little behind the curve, where components are cheaper and better documented? Again, I’d like to hear from you. I’ll post the results of this informal survey in an upcoming editorial.

Backpacks make great things greater. You know, like when Luke carried Yoda around on Dagobah.

The Ardweeny Backpack adds full portability, 5V regulation, Servo, Temperature & Blink-M features to the Arduino-compatible Ardweeny!

PS: We were going to mention broken C-3PO on Chewbacca, but that was more of a mesh than a backpack.
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ADVANCED TECHNOLOGY

WASTEWATER PRODUCES ELECTRICITY AND DESALINATION

There is, of course, a scarcity of clean water for drinking, washing, and other critical uses in many parts of the world, so it would be pretty nice if we could accomplish desalination without relying on reverse osmosis or energy-hogging electrodialysis. It turns out that Bruce Logan, an environmental engineering professor at Penn State (www.psu.edu), along with a cadre of researchers from China's Tsinghua University, have found a way — at least in theory. The team accomplished the feat with a modified microbial fuel cell which uses bacteria to simultaneously clean the water and generate electricity. A typical cell consists of two chambers — one holding the wastewater and the other just plain water — with an electrode in each one. Naturally occurring bacteria in the former consume organic materials and generate a current. The Penn version adds a third chamber in between with ion-specific membranes (i.e., membranes that allow either positive or negative ions — but not both — to pass through) between the central chamber and the electrodes. Salty water is placed in the central chamber. As the bacteria consume the water, negative ions move from the salty water to the wastewater chamber, and positive ions move to the other electrode chamber, thereby desalinating water in the central one. "When we try to use microbial fuel cells to generate electricity, the conductivity of the wastewater is very low," said Logan. "If we could add salt, it would work better. Rather than just add in salt, however, in places where brackish or salt water is already abundant, we could use the process to additionally desalinate salty water, clean the wastewater, and dump it and the resulting salt back into the ocean." Before you get too excited, though, note that there is a catch. It took 200 mL of wastewater (in this case, acetic acid in water) to desalinate 3 mL of saltwater. As Logan observed, "This is not a practical system yet, as it is not optimized, but it is proof of concept." ▲

BYE-BYE BLOODHOUNDS

It may sound a bit morbid, but the National Institute of Standards and Technology (NIST, www.nist.gov) recently demonstrated a new technique for locating grave sites. In the past, cadaver-sniffing dogs or ground penetrating radar has been used for such things, but neither is all that effective at finding bodies that are hidden by concrete or otherwise disguised. As described by NIST, the new process "uses an alumina-coated porous-layer open tubular (PLOT) column with a motorized pipette that pulls in air samples at ambient temperatures. The device detects trace amounts of ninhydrin-reactive nitrogen (NRR) that collects in air pockets above and close to grave soil. Previously, this process involved the tedious and expensive process of solvent extraction of soil samples. Now, a simple probe slightly thicker than a human hair can be inserted into the ground to detect decaying flesh." If you need to locate a stiff under a concrete floor, all you have to do is drill a 1/8 inch hole for probe insertion. To confirm the tester's efficacy, two NIST chemists buried dead rats and took samples over a period of 20 weeks, at which time the decomposing rodents were still detectable. No word yet as to whether they're going to look for Jimmy Hoffa. ▲

COMPUTERS AND NETWORKING

CUSTOM GAMING COMPUTER

Late last year, some former executives of Alienware got together and formed Origin PC (www.originpc.com) with the aim of assembling "the best parts and technology the computer gaming industry has to offer." The result is the Genesis line of highly customized laptop, desktop, and 3D machines that you "build" from the ground up. The first job is to choose the case you want which will run you $189 to
$299, depending on your desired levels of cooling, noise, and upgradeability. Next, you pick the paint job which can push the price of the empty box up to as much as $924. From there, you choose pretty much any desired combination of processors (Intel Core i7 or i5, AMD Phenom II), memory devices, cooling techniques, and other items. After getting past the shock of the box price, things get financially less frightening. For example, if you ordered a machine with a single i7 980X 3.33 GHz processor, an ASUS Rampage III motherboard, liquid CPU cooling, 6 GB of memory, one ATI Radeon HD 5770 graphics adapter, and a pair of 1 TB drives, you'd top out at $4,215 (not including a display, keyboard, and mouse). That's a pretty significant chunk of change, but serious gaming geeks have been known to shell out much more. (For example, check out the Hardcore Reactor X at www.hardcorecomputer.com; it can run you close to $10,000.)

DONATE YOUR SPARE CYCLES

If you're like most of us, you have a lot of computing power that sits idle most of the time. Maybe it's in the form of a PC that you've retired but haven't sold because, frankly, it isn't worth the cost of a classified ad. Or, maybe it's that 3.3 GHz multicore processor in your desktop machine that isn't exactly melting down when it fetches your daily spam. Either way, you might want to consider volunteering a bit of that processing power to a good cause. A good place to start is the University of California, Berkeley's site for volunteer and grid computing, boinc.berkeley.edu. In case you haven't heard of it, BOINC stands for Berkeley Open Infrastructure for Network Computing, and it's used for distributed computing in a range of diverse projects including earthquake detection, scientific and medical research, mapping of the universe, and so on. One interesting project is Einstein@Home (einsteinathome.org) which uses your machine to search for gravitational waves from pulsars, using data from the Laser Interferometer Gravitational-Wave Observatory (LIGO) detector. In fact, a couple months ago, volunteers from Iowa and Germany found one about 17,000 light years away. The current 500,000+ participating BOINC computers grind out almost 2,370 teraFLOPs every 24 hours, so you may as well join in the fun. All you have to do is log onto the BOINC site, choose a worthy project, and download the software. It's a rare opportunity to help create a better world while doing absolutely nothing.

CIRCUITs AND DEVICES

**DVD PLAYER WITH FLO TV**

Most of us are eagerly awaiting the day when we can watch TV without paying exorbitant rates for cable or satellite services, and Audiovox (www.audiovox.com) has taken a step in that direction with its new portable DVD player with FLO TV™. The model DFL 710 offers the usual ability to play DVDs on its seven inch diagonal screen, and it includes stereo speakers. You can also get live mobile television (i.e., real-time transmission rather than downloading or buffering) via Qualcomm's FLO TV service, delivered over "America's largest dedicated mobile TV network." This brings in children's programming (Disney, Nickelodeon), live sports, news (Fox News Channel, CNN Mobile), and a range of other shows (full listing guide at www.flotv.com). The unit lists at $199.99 and, if you grab one quickly enough, includes three months of FLO. Thereafter, you'll need to shell out $14.99/month.

**MEMORY FEATURES WATER COOLING**

You can spend big bucks for a custom, overclocked computer like the ones mentioned previously, but if you have the time and inclination, it's also possible to build one yourself. In that case, the new HyperX memory modules may be of interest. Kingston Technology (www.kingston.com) recently introduced the HyperX H20 water-cooled DDR3 memory kits which offer frequencies up to 2,133 MHz and capacities up to 6 GB. The new line consists of three products: two 4 GB dual-channel kits...
(2,000 and 2,133 MHz) and a 6 GB triple-channel kit (2,000 MHz). According to Kingston’s senior tech manager Mark Tekunoff, “Water cooling is desirable for its quiet operation and long-term reliability. We are bringing HyperX H2O to market as a solution for PC enthusiasts who want to build water-cooled systems.” The units will run you $157, $205, or $235.

INDUSTRY AND THE PROFESSION
2.5 IMAGES COMING BY CHRISTMAS

Back in the 1960s, Ray Dolby introduced the concept of enhanced audio based on reducing noise and enhancing the desired part of the spectrum. Now Paul Darbee, CEO of DarbeeVision, Inc., has come up with a similar approach to image enhancement dubbed DARBEY Visual Presence (DVP). The process is based on experiments Darbee made using stereo videos with dual cameras. Apparently, if you defocus one image and subtract it from the remaining sharp one, you end up with a combined image that produces something that approximates a 3-D viewing experience (he refers to it as

GIZMOS THAT BRING YOUR DREAMS TO LIFE!

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<tr>
<td>3 watt LED strip</td>
<td>$10.00</td>
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<tr>
<td>Noritake Vacuum Fluorescent Display</td>
<td>$20.00 each</td>
</tr>
<tr>
<td>Atmega 8/16/32</td>
<td>$35</td>
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<tr>
<td>Atmel STK500 USB ISP programmer</td>
<td>$38</td>
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Built-In USB, I2C, RS-232 and More
Source Level Debugging

See our full range of products, including books, accessories, and components at:
www.melabs.com
Although the result looks a bit like applying a sharpening filter in Photoshop, it goes beyond that because his images are modified so as to create highlights which your brain interprets as depth. It's a little more complicated than that, so if you want a full explanation, you'll need to visit www.darbeevision.com. In any event, watch for DVP to start appearing in TVs, video game consoles, digital picture frames, and so on by the end of the year. In addition, the company intends to produce "smart" HDMI cables that allow you to retrofit your current high-def TV.

**ARM CONFERENCE NOVEMBER 9-11**

The ARM architecture accounts for something like 90 percent of all low power embedded 32-bit RISC processors which are predominant in PDAs, cell phones, music players, and so forth. If you are involved in designing for this architecture — hardware or software — you should know about the 2010 ARM Technology Conference, scheduled for November 9-11 at the Santa Clara Convention Center. November 9 focuses on chip design issues, and November 10 and 11 are devoted to software and systems design. The conference combines classroom sessions, tutorials, panels, and product demos/exhibits. For more information, visit www.eetimes.com/armconference.
THUMBS UP FOR THE VINCULUM-II TOOLCHAIN

The Vinculum-II embedded dual USB host controller takes all of the hard work out of interfacing and controlling USB devices. In this case the Vinculum-II goes one step further to save us both money and time. We don’t have to design and fabricate a specialized printed circuit board (PCB) to put the Vinculum-II host controller to work. For about $15 plus shipping, you and I can obtain the embedded dual USB host controller IC and all of its associated resistors, capacitors, ferrite beads, and interface pins mounted on a spiffy factory-grade PCB.

A VINCULUM-II EXPANSION BOARD

Odds are that your Vinculum-II project thoughts aren’t exactly the same as mine or the reader in the next town or the reader down the street. In that the FTDI folks have provided an inexpensive Vinculum-II carrier board for its 32-, 48-, and 64-pin parts, taking the time and expense to design and manufacture a generic Vinculum-II embedded dual USB host controller carrier PCB would be nothing more than reinventing the development board wheel.

I just happen to have a 64-pin variant of the Vinculum-II expansion board. The default logical and physical layout of it is drawn up in Schematic 1. The hardware is under the lens in Photo 1. The expansion board interface pins are arranged to fit perfectly into any 0.1 inch pitch perfboard. Thus, we can literally plug our expansion board into a relatively inexpensive perfboard-based embedded dual USB host controller design. No unique PCB is required.

ENOUGH OF THE VINCULUM-II HARDWARE ALREADY

Up to this point, our Vinculum-II discussion has revolved around the hardware and the theory that stands behind it. If you’ve ever worked with any computing device, you know that understanding the hardware design is just as important as understanding how to talk to the intelligent hardware components via firmware. With that thought in mind, let’s apply what we’ve learned about the Vinculum-II hardware in past Design Cycle discussions to writing some Vinculum-II firmware.

THE VINCULUM-II TOOLCHAIN

Like any other toolchain support product, the Vinculum-II toolchain begins life after a successful installation on a PC hard drive. The mainstay of our toolchain is the Vinculum-II IDE. Like most other IDEs (Integrated Development Environments), the Vinculum-II IDE allows the programmer to create projects, edit files, compile code, program devices, and debug application code.

The power behind this toolchain lies in the code that forms the FTDI libraries. The FTDI libraries consist of device drivers, runtime libraries, and the VOS Kernel Services. Some of you are familiar with what device drivers and runtime libraries contribute to your code. However, you may not be privy to what the Vinculum-II VOS Kernel Services have to offer. The services provide overall application control while supplying the necessary primitives and unique data structures needed to support Vinculum-II applications.

The Vinculum-II toolchain uses an API (Application Program Interface) to access the functionality of the toolchain’s device drivers and Kernel services. The API is under the control of the Device Manager which is under the command of the Kernel Services. The inclusion of a Vinculum-II API makes most of the pain out of application development.

PHOTO 1. The Vinculum-II carrier board sits on pins that are all fitted at 0.1 inch centers which makes this carrier board perfect for integrating with standard 0.1 inch pitch perfboards.
as the language behind the calls to the functions that reside within the device drivers is simplified. The API structure is also designed to provide the programmer easy access to the VOS Kernel Services in the same way it does for the device drivers. In most cases, the API function call names describe their actual intended function. Thus, the API calls are self-commenting. I don’t think I have to say a word about what these API calls do:

vos_dev_write
vos_dev_read
vos_delay_msecs
vos_dev_open
vos_dev_close
gpio_init
uart_init

**TYPICAL VINCLUDUM-II APPLICATION COMPONENTS**

The C programming language and its syntax are used to build Vinculum-II applications. These applications usually contain a user-generated application header file, a user-generated main function, and user-generated threads. Application header files are not a requirement in Vinculum-II application code. However, for more complex applications, user-generated application header files are recommended as they provide a common holding point for the application’s global variables and constant values.

For those of you that can C already, you know that all of the initial application action is directed from the main function. In the case of the Vinculum-II, the threads contain the actual application functionality. The main function contains code that reads the system before starting the threads. The very last thing a Vinculum-II main function does is start the scheduler which starts the execution of the threads.

You can get a pretty good idea of the available Vinculum-II hardware resources by examining ScreenShot 1. As you can see, the toolchain has built-in device driver header files for all of the on-chip peripheral components. There are device driver headers for file management, general-purpose I/O, SPI, timers, UARTs, and USB.

The available toolchain device driver libraries are captured in ScreenShot 2. For the most part, there are name-associated device driver libraries that work hand in hand with the like-named device driver header files shown in ScreenShot 1. You can add any of these header files to your project and inspect them with the IDE editor. However, you can’t gain view access to the device driver...
libraries via the Vinculum-II IDE editor.

Screenshot 3 reveals the VOS Kernel Services header files while Screenshot 4 scrolls down to the Runtime headers. As you have most likely already concluded, there are library files that are associated with the header files and they can be seen in Screenshot 5. If a library is included in an application, its associated header file must also be included in the application.

NAVIGATING VINCULUM-II APPLICATION SOURCE CODE

VOS Kernel Services, device drivers, and API calls are major parts of the Vinculum-II application code big picture. However, we can’t write an application by simply including libraries and headers into our Vinculum-II application source code. So, let’s practice what we preach and begin by coding up an initial Vinculum-II application header file.

INSIDE THE APPLICATION HEADER

The application programmer guide application note states that the first thing we should include in our application header file is the size of the stack memory that the application thread will require:

```
#define SIZEOF_FIRMWARE_TASK_MEMORY 0x1000
```

The stack memory size of 0x1000 is overkill by design. After successfully compiling, loading, and running the application we’re about to discuss and build with the 0x1000 value, I was able to get the application to run reliably with a stack size of only 0x0400. Thus, the stack memory size value is dependent on the complexity of your application.

The next recommended programming action is to fix a number of devices that the application will use:

```
#define NUMBER_OF_DEVICES 5
#define VOS_DEV_USB_HOST1 0
#define VOS_DEV_BOMS 1
#define VOS_DEV_USB_HOST2 2
#define VOS_DEV_UART 3
#define VOS_DEV_GPIO 4
```

Each device we list must have a unique device identifier which is used later by the Device Manager. If you’re having problems with the pair of HOST definitions, recall that the Vinculum-II IC has a pair of USB portals. These portals can be identified in Schematic 1 as USB1 and USB2. The last entry in our application header file is a forward declaration to the user-generated application thread which is really no more than a C function. A forward declaration is coded for each thread in the application. In our case, we only have one:

```
void application_thread(void);
```

With the application-specific header file code completed, we can move on into the application’s source code file which we will call DesignCycle-App.c. Naturally, the application header file we just coded is named DesignCycle-App.h. Our first order of business is standard C fare. We will offer sacrifices unto the #include gods:

```
#include "vos.h"
#include "devman.h"
#include "DMA.h"
#include "STM32H7.h"
```

You will recognize the aforerecorded #include statement arguments as members of the resident Kernel header files (see Screenshot 3). The VOS Kernel Services header file vos.h is a must-have as it supports the Kernel library which is the core power behind our application. Our application #include list acts as a guide for the device driver files we
need to include in the application source code space:

```c
#include "USBHost.h"
#include "USB.h"
#include "MST.h"
#include "BOMS.h"
#include "UART.h"
#include "FAT.h"
#include "GPIO.h"
#include "string.h"
#include "DesignCycle-App.h"
```

The `MST.h` and `FAT.h` files support the `BOMS.h` functionality. `MST` is short for Mass Storage Interface while `FAT` is the good old Microsoft acronym that translates to File Allocation Table. `BOMS` — Bulk Only Mass Storage — is a USB class that describes a device for communicating with mass storage devices. In the Vinculum-II world, a mass storage device doesn't physically get any bigger than a thumb drive.

Once a device is opened, Device Manager returns a unique handle for that device. The returned handle is of the type `VOS_HANDLE`. Here's how `VOS_HANDLE` is declared within the Kernel's `devman.h` file:

```c
#define VOS_HANDLE uint16
VOS_HANDLE vos_dev_open(uint8 dev_num);
```

`VOS_HANDLE` is simply an unsigned 16-bit integer that holds the unique handle value returned by the `vos_dev_open` function which is one of the Device Manager functions. According to the device list coded in the application header, we need to reserve five unique VOS_HANDLE slots:

```c
VOS_HANDLE hUsb1, hUsb2, hUart, hBoms, hGpio;
```

The FAT driver needs to have a context declared to allow it to communicate with the file system on our mass storage device. Here's how the FTDI folks do it:

```c
fat_context fatContext;

And, that's how we will do it. The `fat_context` code that follows is found within the `FAT.h` header file. The FAT driver is layered on top of the `BOMS` driver. Thus, the context declaration is a bit different from context code you'll be exposed to as we continue:

```c
  // context pointer for instance of FAT
  // file system
typedef void *fat_context;

  fat_context *fat_open(VOS_HANDLE hBoms, unsigned char partition, unsigned char *status);

  According to the `fat_context` source code we can gather from the various Vinculum-II toolchain header files, it seems that the FAT context data is derived from the invocation of the `fat_open` function. The `fat_close` function adds fuel to that fire as the returned `fat_context` information is also used by the `fat_close` function:
```c

  void fat_close(fat_context *fat_ctx);

  The `vos_create_thread` function returns a pointer to the newly registered thread. So, we'll need to accommodate that action by declaring a pointer to our user-generated application thread which we will name `application_thread`:
```c

  vos_tcb_t *tcbApplication_thread;

  Although the thread is not created until the last moment in the main function, I'll provide an advanced look at the thread creation function call so you can relate the `tcbApplication_thread` pointer we declared to the thread creation process:
```c

  tcbApplication_thread = vos_create_thread
(29, SIZEOF_FIRMWARE_TASK_MEMORY,
 application_thread, 0);

  The first parameter in the `vos_create_thread` function
```c

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void main(void)
{
    // USB Host configuration context
    usbhost_context_t usb_ctx;

    // Context for UART
typedef struct __uart_context_t {
        unsigned char buffer_size;
    } uart_context_t;

    An instance of the uart_context_t structure called
    uart_ctx is created within the confines of the main
    function in this manner:

    // UART configuration context
    uart_context_t uart_ctx;

    I think you can get there from here as far as the GPIO
    context is concerned. Here's the GPIO seed context
    structure definition:

    // GPIO context
typedef struct __gpio_context_t {
        unsigned char port_identifier;
    } gpio_context_t;

    And ... here's the code that declares an instance of
    the gpio_context_t structure type:

    // GPIO configuration context
    gpio_context_t gpio_ctx;

    The common element of all of the context structures is
    configuration data that pertains to the device the structure
    supports. Each device context is used to set the
    configuration of a device before it is opened.

    Now that we've voiced our intentions to support, configure,
    and possibly run five devices, let's prime the pump by initializing
    the VOS Kernel Services. For clarity, I've taken the liberty to
    pull the vos_init parameter values from their respective
    header files and place them in the vos_init function call:

    #define NUMBER_OF_DEVICES 5

    //***************************************************************
    // VOS INITIALIZATION
    //***************************************************************

    // VOS initialization and start-up
    void vos_init(uint8 quantum, uint16 tick_cnt, uint8 num_devices);
    void vos_start_scheduler(void);

    // default interval for timer interrupts
    #define VOS_TICK_INTERVAL 1

    // default time-slice quantum for tasks in
    // RUNNING state
    #define VOS_QUANTUM 50

    vos_init(10, VOS_TICK_INTERVAL, NUMBER_OF_DEVICES);

    The VOS_QUANTUM default value has been

---

If you're an avid Design Cycle reader, this USB host code
is no stranger to you. The typedef struct source code snippet
defines a structure of type usbhost_context_t. We can create
instances of the structure type usbhost_context_t like this:

---

**SCREENSHOT 5.** Every header file that is directly
associated with a like-named library file contains
definitions and data structures to support the functions
contained within the library file it supports.

(29) is the thread priority. A priority of 31 is highest with a
priority of one being lowest. Recall that we declared the
sizeof_FIRMWARE_TASK_MEMORY value in our application
header file. We also coded a forward reference to the
user-generated application thread (application_thread) we
are creating. The thread application_thread is not coded to
accept any arguments which explains the zero at the end of
the thread creation function.

---

**INSIDE THE MAIN FUNCTION**

It's time to put all of those plans we made in the
application header file into action. Let's begin by declaring
a context for the USB host. Before we actually code the
host context declarations, let's examine the code structure
behind a USB host context. All we have to do is pull up the
USBHost.h file in the Vinculum-II IDE:

// Context for USB Host
typedef struct __usbhost_context_t {
    // number of interfaces both USB
    // hosts combined
    unsigned char if_count;
    // number of endpoints (excluding control
    // endpoints) expected
    unsigned char ep_count;
    // number of concurrent transaction
    // expected
    unsigned char xfer_count;
    // number of concurrent isochronous
    // transactions expected
    unsigned char iso_xfer_count;
} usbhost_context_t;

---

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overridden with a value of 10 decimal. The VOS_xxxx definitions were gleaned from the vos.h header file. We entered the NUMBER_OF_DEVICES value when we coded the application header file DesignCycle-App.h.

This would be a good time to set up the Vinculum-II clocking. As you might imagine, the toolchain API has a call for that:

```c
vos_set_clock_frequency(VOS_48MHZ_CLOCK_FREQUENCY);
```

As you can see in Schematic 1, the 48 MHz clock is derived from a 12 MHz crystal and the Vinculum-II’s 4x PLL. In a previous Design Cycle Vinculum-II discussion, we took a detailed look at a helper program called the VNC2 IOMUX Config utility which is part of the Vinculum-II toolchain package. If you had the opportunity to join in on that conversation, you’ll recall that the VNC2 IOMUX Config utility is used to configure the GPIO subsystem. An added feature of this utility is that it writes the GPIO configuration code for us. Well, we need some UART I/O code for sure and if we want to add some status LEDs to the mix, we should go ahead and lay out some output I/O pins to support them, as well. Here’s what the VNC2 IOMUX Config utility produced according to my idea of where things should go:

```c
// GPIO port A bit 2 to pin 13 - LED4 ON // EVAL BOARD
vos_iomux_define_output(13, IOMUX_OUT_GPIO_PORT_A_2);
// GPIO port A bit 5 to pin 29 - LED5 ON EVAL // BOARD
vos_iomux_define_output(29, IOMUX_OUT_GPIO_PORT_A_5);
// GPIO port A bit 6 to pin 31 - LED6 ON EVAL // BOARD
vos_iomux_define_output(31, IOMUX_OUT_GPIO_PORT_A_6);
// UART to V2EVAL board pins
vos_iomux_define_output(39, IOMUX_OUT_UART_TXD);
vos_iomux_define_output(40, IOMUX_OUT_UART_RXD);
vos_iomux_define_output(41, IOMUX_OUT_UART_RTS_N);
vos_iomux_define_output(42, IOMUX_OUT_UART_CTS_N);
```

I’m sure you’re wondering why I chose these particular I/O pins and LED identifiers. The cat is let out of the bag in the comments area of the code generated by the VNC2 IOMUX Config utility. I guess it would be a good idea to add some LED definitions to our application header code:

```c
#define LED0 0x02 // LED3 ON EVAL BOARD
#define LED1 0x04 // LED4 ON EVAL BOARD
#define LED2 0x20 // LED5 ON EVAL BOARD
#define LED3 0x40 // LED6 ON EVAL BOARD
```

It is impossible for us to produce a single PCB and purchase all of the necessary electronic components to

---

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our board mounted on a Vinculum-II Evaluation Board.

Thus far, all of our Vinculum-II application code has targeted the Evaluation Board peripheral layout. With that, let's move on and continue bringing the Board peripherals to life. The USBHost context contains a field that holds the maximum number of interfaces to support (if_count). We'll use the if_count value and the rest of the USBHost context to initialize the USBHost interfaces:

```c
usb_ctx.if_count = 2;
usbhost_init(VOS_DEV_USB_HOST1, VOS_DEV_USB_HOST2, &usb_ctx);
```

The following sequence of events wraps up the actions initiated inside of the main function. The BOMS service is initialized, a UART buffer is allocated, the UART is initialized, the GPIO port is initialized, the application thread is created, and the scheduler is started which allows the thread code to be executed:

```c
boms_init(VOS_DEV_BOMS);

uart_ctx.buffer_size = VOS_BUFFER_SIZE_128_BYTES;
uart_init(VOS_DEV_UART, &uart_ctx);
gpioCtx.port_identifier = GPIO_PORT_A;
gpio_init(VOS_DEV_GPIO, &gpioCtx);

tcbApplication_thread = vos_create_thread(29, SIZEOF_FIRMWARE_TASK_MEMORY, application_thread, 0);

vos_start_scheduler();

main_loop:
goto main_loop;
```

The scheduler will continually queue the application thread for execution as the main function turns in a very tight loop.

### Things Associated with Application Threads

Lots of device open and ioctl (I/O Control Block) operations are the norm for an application thread. For instance, here is the sequence of events to set up the GPIO to support the LEDs:

```c
// THIS CODE FOUND IN GPIO.h
#define VOS_IOCTL_GPIO_SET_MASK 1

// set pins to either input (0) or output (1)

// GPIO control block for use with GPIO IOCTL
```
I pulled the `SET_MASK` and GPIO control block definitions from the `GPIO.h` file and added them in the code mix for clarity. Once the `gpio_ioctl` control block is instantiated, it gets used as a tool to set the I/O data direction. Just before the I/O pin data direction is set, the GPIO device is opened and its handle is returned to `hGPIO`. The Vinculum-II Evaluation Board LEDs are all connected as common anode with respect to the power source. Thus, the Vinculum-II drives the cathodes from its I/O pins. These lines of code will illuminate all of the LEDs we’ve defined:

```c
leds = 0;
vos_dev_write(hGPIO, &leds, 1, NULL);
```

These lines of code will extinguish all of the defined LEDs:

```c
leds = 0xF0;
vos_dev_write(hGPIO, &leds, 1, NULL);
```

Want to turn off LED0 only? Here’s how we do it:

```c
leds = LED0;
vos_dev_write(hGPIO, &leds, 1, NULL);
```

The `vos_dev_write` API call uses the GPIO handle (`hGPIO`) to identify the GPIO pins that are driving the LEDs and writes the value of `leds` to the I/O port. Pretty clever, huh?

---

**THUMB DRIVE HEAVEN**

Communicating with a thumb drive is just as easy as controlling the bank of LEDs. The FTDI Vinculum-II site contains detailed code examples of how to put a thumb drive online, as well as the Vinculum-II toolchain and a treasure trove of application notes. You’ve all been asking me to deliver low cost projects with a punch. I think I have succeeded. For about 100 bucks you can go to thumb drive heaven and add the Vinculum-II embedded dual USB host controller to your Design Cycle.  

---

**SOURCES**

FTDI  
Vinculum-II Embedded Dual USB  
Host Controller  
Vinculum-II IDE  
Vinculum-II Toolchain  
www.ftdichip.com
Blinky-Eyes Animated Display
- Animated display of 66 super bright LED's!
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The ultimate animated LED kit that will dazzle you and delight your friends! Uses a microcontroller to randomly select from many different animations such as a long pause before a wink, or a twinkle of the eye to startle passers-by!

Four modes to satisfy any enthusiast’s desires: 1. Off for long random periods, then blinks or winks. Designed to scare! 2. On for long periods before performing an animation, perfect for costumes and displays! 3. Animates all the time for constant motion, perfect display attention-getter. 4. Random Fire! When placed in a pumpkin light it up like you wouldn’t believe! As if this weren’t enough, the B66 can also control a small hobby motor to shake bushes at random intervals or signal an external player to make a scary sound! Also has a CDS cell to sense light. In one mode, the display will dim as it gets dark, for battery operation, and in the other it will turn off when it’s too bright, so it plays only in the dark!

BE66  Blinky-Eyes Animated Display Kit $59.95

Automated Animated Ghost
- Automatically greets your visitors!
- Responds to sudden noises!
- Built-in microphone!
- Adjustable sensitivity

Who says ghosts are make believe? Once a friend or indeed any one you think they might think differently! The unique circuit board design includes two omenous blinking eyes that change with various conditions, including sudden changes in ambient noise. A highly sensitive built-in microphone picks up anything from noises to talking and makes the ghost dance with its built-in motor, make eerie sounds with the built-in speaker, and randomly blink. A white cloth and a hanger are included as shown to make it look like the real thing. Runs on 2 AAA batteries (Not included).

MK166  Automatic Animated Ghost Kit $21.95

Halloween Pumpkin
- 25 bright LED’s!
- Random flash simulates flickering candle!
- Super bright LED illuminates entire pumpkin!
- Simple & safe 9V battery operation

The perfect “starter” kit with a terrific Halloween theme! You won’t be scraping the seeds and guts out of this pumpkin! Six transistor circuit provides a neat random flash pattern that looks like a flickering candle. Then a super bright LED illuminates the entire pumpkin with a spooky glow!

The pumpkin face is the actual PC board, and assembly is easy through-hole soldering of all components and LEDs. Your pumpkin is powered by a standard 9V battery (not included) which snaps to the back of the pumpkin. An on/off switch is also included. Create a new kind of pumpkin this year, and learn about LED’s and electronics at the same time!

MK145  Electronic Halloween Pumpkin Kit $10.95

Portable EL Electroluminescence
- 3.3 feet long!
- Low power consumption!
- Highly visible brilliant colors

Electroluminescence illuminated flexible wire sets can be used for a lot of things but when they’re battery powered they’re perfect for Halloween and Trick or Treat safety concerns! Each thin, flexible EL wire set is 3.3’ long and runs on two standard AA batteries (not included). Current consumption is less than 100mA for long life.

Mode settings include steady glow and slow or fast flash! Make it part of a brilliantly lit, custom designed costume of simply add it for illuminated safety while Trick or Treating in the dark.

NWR15  EL Illumination Wire Set, Red $19.95
NWR15  EL Illumination Wire Set, Green $19.95
NWR15  EL Illumination Wire Set, Blue $19.95

Spark Generating HV Plasma Generator
- Generate 2” sparks to a handheld screwdriver!
- Light fluorescent tubes without wires!
- Build your own Plasma Balls!
- Generates up to 25KV @ 20 kHz from a solid state circuit!

This popular kit was conceived by one of our engineers who likes to play with things that can generate large, loud sparks, and other frightening devices! And at Halloween there’s no better effect than high voltage sparks flying through the air! The PG13 Plasma Generator creates a very impressive 25,000 volts at 20 kHz, to provide a stunning display of high voltage! It will draw a cool looking 2” spark to hand held screwdriver, or light fluorescent tubes without any connection!

It produces stunning lighting displays, drawing big sparks, to perform lots of high voltage experiments. In the picture, we took a regular 2” “Decor” style light bulb and connected it to the PG13 - WOW!!! A storm of sparks, light traces and plasma filled the bulb. Holding your hand on the bulb doesn’t hurt a bit and you can control the discharge! It can also be used for powering other experiments; let your imagination be your guide! Operates on 16VAC/VDC for maximum output. Can also be run from 5-16VAC/VDC to reduce the output voltage.

PG13  Plasma Generator Kit $64.95
PS21  110VAC Input, 16VAC Output, Power Supply $19.95

Tri-Field Meter & “Ghost Detector”
- See electric, magnetic, and RF fields!
- Watch the magnetic fields of the earth!
- Sense different magnetic poles
- Detect RF transmitter fields
- Graphical LED display allows you to “See” the invisible fields
- Great learning tool for EMF, RF, and man-made field theory.

Call it a Tri-Field Meter, an Electrical, Magnetic, and RF Detector, a Ghost Detector, or a Tricorder, but walking around with this on Halloween will seriously impress even the most doubting!

The TFM3C has three separate field sensors that are user selectable to provide a really cool readout on two highly graphical LED bargraphs! Utilizing the latest technology, including Hall Effect sensors, you can walk around your house and actually “SEE” these fields around you! You will be amazed at what you see. How sensitive is it? Well, you can see the magnetic field of the earth. That’s sensitive!

The technical applications are endless. Use it to detect radiation from monitors and TV’s, electrical discharges from appliances, RF emissions from unknown or hidden transmitters and RF sources, and a whole lot more! If you’re wondering whether your wireless project or even your cell phone is working, you can easily check for RF. A 3-position switch in the center allows you to select electric, magnetic, or RF fields. A front panel “zero adjust” allows you to set the sensors and displays to a known clean “starting point.”

If the TFM3C looks familiar, it’s probably because you saw it in use on the CBS show Ghost Whisperer! It was used throughout one episode (#78, 02-27-2009) to detect the presence of ghosts!

The concept is simple, it is believed (by the believers) that ghosts give off an electric field that can be detected with the appropriate equipment. In the electric mode, the TFM3C’s displays will wander away from zero even though there isn’t a clear reason for it (not scientifically explainable, aka paranormal). This would mean something has begun to give off an electric field. What it was in the Ghost Whisperer was a friendly ghost. What it will be in your house...who knows! Makes a GREAT learning project besides! Requires 4 AA batteries.

TFM3C  Tri-Field Meter Kit With Case $74.95
OBII CarChip Pro
The incredible OBII plug-in monitor that has everyone talking! Once plugged into your vehicle it monitors up to 200 hours of data from speed, braking, acceleration, RPM and a whole lot more. Reads and resets your check engine light, and more!

8226 CarChip Pro OBII Monitor $99.95

Practice Guitar Amp & DI
Practice your guitar without driving your family or neighbors nuts! Works with any electric, acoustic-electric, or bass guitar. Plug your leads into the input section for the perfect preamp practice to your favorite music! Drives standard headphones and also works as a great DI.

PG1 Personal Practice Guitar Amp Kit $64.95

Passive Aircraft Monitor
The hit of the decade! Our patented receiver hears the entire aircraft band without any tuning! Passive design no IC, therefore can be used on board an aircraft! Perfect for airshows, hears the active traffic as it happens!
Available kit or factory assembled.

ABM1 Passive Aircraft Recv Kit $99.95

Electronic Siren
Exactly duplicates the upward and downward voice of a police siren. Switch closure produces upward whool, releasing it makes it return downward. Produces a loud SW peel and will drive any speaker! Horn speakers sound the best! Runs on 6-12VDC.

SM3 Electronic Siren Kit $7.95

LED Emergency Light
Our #1 Mini Kit for over 35 years! Alternately flashes red LED's. Great for signs, name badges, model railroading, and more. Used throughout the world as the learning aid kit for students young and old! Great practice kit for students.

BL1 LED Blinkly Kit $7.95

Universal Timer
Build a time delay, keep something on for a preset time, provide clock pulses or provide an audio tone, all using the versatile 555 timer chip! Comes with circuit theory and a lot of application ideas and schematics to help you learn the 555 timer. 5-15VDC.

UT5 Universal Timer Kit $9.95

20 Watt Mini Amp
Delivers a super clean 20W output from one SMT package! Ultra efficient circuit design produces no heat. PCB can be snipped into a small circle for special applications. Runs on 18VDC for rated output, or down to 10VDC for reduced output.

UAM2 20W Subminiature Amp Kit $34.95

Touch Switch
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T51 Touch Switch Kit $9.95

Walking Electronic Bug
Built around a pair of subminiature cell phone motors, this bug wanders around looking for things to bump into! Sensors below his LED eyes sense proximity and makes him turn away! Steer him with flashlights too! Runs on two "N" batteries.

WEB1 Walking Bug Kit $29.95

Stereo Ear Super Amplifier
Ultra high gain amp boosts audio 50 times and it does it in stereo with its dual directional stereo microphones! Just plug in your standard earphone or headset and point towards the source. Incredible gain and perfect stereo separation!

MK136 Stereo Ear Amp Kit $9.95

Water Sensor Alarm
This little kit can really "bail you out"! Simply mount the alarm where you want to detect water level problems (sump pump). When the water touches the contacts the alarm goes off! Sensor can be remotely located. Runs on a standard 9V battery.

MK108 Water Sensor Alarm Kit $6.95

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WITH RUSSELL KINCAID

In this column, I answer questions about all aspects of electronics, including computer hardware, software, circuits, electronic theory, troubleshooting, and anything else of interest to the hobbyist. Feel free to participate with your questions, comments, or suggestions. Send all questions and comments to: Q&A@nutsvolts.com

LARGE CLOCK

Q I have read with real interest your articles on the large clock using the PIC. I would like to make a clock that can be programmed with the Arduino and use no surface-mount chips. (It is hard to breadboard with SMD.)

But I want the complete clock ... meaning LEDs for all the hours, all the minutes, and all the seconds. The seconds would consist of a circle of 60 bi-color LEDs (green and red) with the green on all of the time and the red coming on each second in rotation.

The minutes would consist of a circle of 60 yellow LEDs, each of which would stay lit for the minute indicated and go off when the next minute comes on. Also, when the minute LED lights up it needs to trigger a cluster of LEDs that can be formed into the minute hand.

The hours would consist of 12 LEDs that stay on for the hour indicated and also trigger a cluster that can be formed into the hour hand.

In addition, there needs to be a means of setting the time. This would then be a digitally driven analog clock with no motors or moving parts!

Any advice or help would be greatly appreciated.

— Jay Simmons

A I am not going to learn a new programming language and use an unfamiliar chip, but I could design a circuit and PICBASIC program for a PIC16F57. You can convert the program to C++ for the Arduino and I am sure you can find a version among the 11 now produced that will be compatible.

Wait a minute! Sixty lines for the seconds, 60 lines for the minutes, 12 lines for the minute hand, and 12 lines for the hour hands: 144 lines. I don’t know of a micro with 18 eight-bit ports, so it looks like a hardware solution is easier.

Four 74HC154 (one of 16 decoders) will give the 60 lines for the seconds and another four for the minutes. An hour counter will use another 74HC154 to move the hour hand (see Figure 1). I could still use a PIC16F57 to control the 74HC154s but the cost in time far outweighs the cost of a half dozen 30 cent chips.

The seconds are displayed by bi-color LEDs (red and green). When the red LED is turned on by Q1, the green goes off because its voltage drop is higher. Instead of having the minute LED go off, I suggest a circle of 60 yellow LEDs, with every fifth one being red to facilitate reading the time. These LEDs will be on all the time. The minute hand will rotate with 60 positions; the hour hand will have 12 positions. You can have up to 20 LEDs in each hand with the ZVP2106A or, if you want to use more, the NDP6020P will handle up to 2,000.

Here is how the circuit works: The line frequency is divided by 60 to produce one second clocks to IC5 which is a binary counter. This drives IC1, IC2, IC3, and IC4 which represent one of 16 decoders. When the count of IC1 gets to 16, feedback through IC14A resets IC5 to zero, which decodes to one. At the same time, IC15, a synchronous shift register, is clocked which enables IC2 and disables IC1. This process continues until we get to the 16th count in IC4. The output from IC4 through IC17F puts IC15 in parallel entry mode and the clock from IC17D does the parallel entry which enables IC1, and the cycle starts over. The reason for the two inverters, IC17D and IC17E, is that the clock has to happen after the parallel entry mode is set, so the clock is delayed by two inverter time delays. The delay is less than a microsecond so it won’t be noticed. The minute circuit is identical to the second’s circuit, and the hour circuit differs only in that it simply has to count to 12.

To set the clock, the normal operation is disabled and the one second pulses from IC8B are used to advance the count. Switches A, B, C, and D are momentary pushbuttons. Switch A will advance the seconds in 15 second increments and switch B increments in one second increments; you can set the seconds in less than 19 seconds and the minutes the same way. The hours increment one hour per second so that is fast also.

I have not built or simulated this
circuit, so I will appreciate any feedback or concerns about my logic.

**HIGH VOLTAGE SINE WAVE**

I want to generate a sine signal at 10 kHz and also 10,000 volts for a project that I am working on. I have seen the Royer oscillator at http://wiki.4kv.org/index.php/royer_oscillator. This system uses a flyback for the output but I am not sure that it is a sine wave. I also need to make a 25 kHz sine signal at 10,000 volts.

PS: Can you recommend a good electronics dictionary?

I don’t know if the Royer oscillator will be a good sine wave so I am proposing a sine oscillator and linear amplifier (class D) (see Figure 2). You don’t say what the load is on the 10 KV output, but if you want 10 KV at 10 mA, that is 100 watts. If your power requirements are less, you can reduce the size of the power amp and power transformer. I am going to assume that you use the coil FLYPVM400 from Information Unlimited because I know how many turns are on it. If you use some other
coil, you will have to experiment to determine the number of primary turns.

In this case, the amplifier is capable of 100 watts into 8 ohms. The voltage output ($P=V^2/R$) is 28 volts AC. In order to get 10 KV out, you need a turns ratio of 10,000/28 = 357. The primary turns are therefore: 4000/357 = 11. The primary current will be in the order of 357*10 mA = 4 amps, so #14 wire would be good.

I found an amplifier on eBay from China (TDA8920); it’s 60 watts stereo and you can parallel the two amplifiers for 120 watts output. The amplifier requires 20-0-20 VAC or 40VCT; Mouser part number 546-182524 should work. The oscillator circuit is adapted from National Semiconductor Application Note: AN31, Figure 29, Wein bridge oscillator. This is the circuit used by HP in their 200CD oscillator. The 1869 lamp (Mouser part number 606-CM1869) regulates the amplitude because its resistance increases when the signal amplitude increases. The frequency is given by: $F = 1/(2\pi R_5C)$ when $R_4 = R_5$ and $C_1 = C_2$.

It may be out of print, but I find Rudolf F. Graf’s Modern Dictionary of Electronics, published by Howard W. Sams to be useful. There is also Glossary of Communications by

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**OBSCURE BATTERY**

I am having a problem locating a 12 volt SLA battery for a large toy truck. The battery name is DYNAFLO 6-FM-4.0 12 Volt 2.3 Ah; it has a three terminal Molex connector. I can find some similar batteries but none have the three connections. No one can tell me what the three wire configuration is. I am not sure if a 12 volt two terminal SLA could be configured with the three-pin Molex connector.

— Charles Huth

---

**A**

I believe the red and black wires are the positive and negative of the battery. The blue wire is probably connected to a thermistor to measure the temperature of the battery. If you charge another battery without the thermistor, you can protect it from overheating by connecting a resistor between the charger and battery. Measure the charger voltage ($V_c$), then compute the resistor value from: $R = (V_c-13.8)/I$ where $I = 2.3$ amps divided by the charge time. Ten hours is the standard charge time but you could use less if you check the temperature. The power rating of the resistor is $I^2R$.

---

**Q**

I am trying to find a low cost and easy to use piece of software to simulate circuits. Back in 2008, my high school teacher recommended a program called Circuit Shop and it worked out well for a little more than a year; then I started learning and designing more complex circuits and the software just couldn’t keep up. It would return answers on capacitors and inductors that I didn’t even have to calculate to know they were wrong. I had a friend just a few months ago recommend LTSpice saying it is what his college uses but warned me that it was less than user friendly. After messing around with it for a little over a month now, I can see what he means. I have to navigate through two or three different windows just to set up the parameters on a resistor. Currently, my college uses MultiSim but that is way out of my price range. Can you recommend some really good circuit simulation programs that are fairly easy to use, reflect what I might see when I get a job in the industry, and won’t break the bank?

— Cameron Seidi

**A**

Any circuit simulator is going to involve a learning curve. You need to read the user’s manual available at: [http://ltspice.linear.com/software/scad3.pdf](http://ltspice.linear.com/software/scad3.pdf). There is also a user’s forum in Yahoo Groups: [http://groups.yahoo.com/group/LTspice](http://groups.yahoo.com/group/LTspice). I have used Ispice, IGSpice, Pspice, Electronic Workbench, and am presently using LTspice. LTspice is full functioning freeware and I have not found any program like it. I have Tina-TI from Texas Instruments but
when I tried to use it for a passive circuit, I got an error message: No IC in the circuit. I did a Google search and found 5spice which looks to be easy to use but costs $199. There is a free demo but it is not useful. I also found Opus spice which is free but I have not evaluated it or figured out how to install it.

LTspice has made some modifications that minimize “time step too small” errors in switching circuits so it is good for those types of circuits. In any simulator, the results are only as good as the model, but the more complex the model, the more likely that the program will fail. Always use the simplest model that will give the results you need and keep in mind that if you use an IC in an atypical way, it may not work because of assumptions that the model designer made.

BLACK BODY HEAT SOURCE

I need to calibrate some infrared thermometers. I hacked an old black body heat source, modifying the well to accommodate the IR thermometers and an NIST traceable RTD thermometer. My issue is that the source only produces a single (100 deg C) temperature. My goal is to be able to construct a stable multiple point heat source (e.g., 50, 100, 150 deg C) or a variable source of the same range.

Here are some particulars about the old black body heat source, some of which I want to reuse: The heat is provided by a coil of resistive wire which measures 9.5 ohms at 75 deg F, which is wound around a copper bobbin with a well painted flat black. It is powered by a 12V 7 AH SLA battery (nice because it makes it quite portable) and it uses a temperature sensor which measures ~800 ohms at 75 deg F, and uses a uA78S40PC IC to do the control.

The only components that I want to reuse are the well with the heating coil and the 12V power source.

Purchased heat sources are mega bucks and I'd like to do this cheaper. Would you please help me?
— Mark Hoffman

A

Since the 78S40 has been doing the job, we may as well continue with that. I don't know how much power is required but with 12 volts and 9.5 ohms, it can't be much more than 10 watts. I expect the power to be much less than that if it is well insulated. I will use the step down circuit of the datasheet and calculate the parameters using the Design Formulas from the datasheet.

Vo = 10V (that's the most we can get with 12V supply)
Vd = 1.5V (diode drop at 1 amp)
VR = 12V (battery)
Vsat = 1.3V (transistor drop at 1 amp)
Ton/Toff = 12 (from datasheet formula)
Set Fmin to 20Khz (so you can't hear it)

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October 2010 NUTSIVOLTS 27
You will be checking the temperature with your RTD but for temperature feedback, I will use LM35—a Celsius temperature sensor. The output is 10 mV per degree C, so the output will be 1.5V at 150 deg C. I want to be able to read the set temperature with a digital voltmeter, so I can't use the internal reference of 1.25V. The LM4040-2 (2.048 volt reference) will be able to provide 1.5 volts to set the 150 deg C temperature (see Figure 3).

The LM35 should be clamped to the well because the initial temperature overshoot will be proportional to the thermal resistance between the well and the sensor. The TO-46 metal case would be best but I did not find any; the part is plastic (TO-92). I recommend connecting C4 directly to the LM35 pins and use #36 wire to connect to the outside of the insulation to minimize conducted thermal energy. R5 is a 10 turn pot, so you can set the temperature more accurately. NV

---

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- Kit supplied with PCB, speaker, socket and electronic components.
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- PCB: 105 x 60mm

Instructions NOT included. See KI-8091 $1.25 for individual instructions or full color project book BJ-8505 $7.50

Instructions NOT included. See KI-8091 $1.25 for individual instructions or full color project book BJ-8505 $7.50

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KC-5417 $20.50 plus postage & packing
Enables you to drive up to two stereo headphones from any line level (1 volt peak to peak) input. The circuit features a facility to drive headphones with impedances from about 8-600 ohms. The Jaycar kit comes with all specified board components and quality fibreglass tiled PCB.
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- PCB board size: 134 x 103mm

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KC-5152 $4.75 plus postage & packing
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KC-5172 $9.50 plus postage & packing
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- PCB: 47 x 44mm

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- PCB: 165 x 60mm

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- PCB: 120 x 80mm

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KC-5496 $20.00 plus postage & packing
Remote controlled digital timer with a bright 20mm high 7-segment red LED display. It can count up or down from one second to 100 hours in 1-second increments. Its timing period can either be set and controlled using the remote control or it can be automatically controlled via external trigger/reset inputs. An internal relay and buzzer activate when the timer times out. The relay contacts can be used to switch devices rated up to 30VDC or 24VAC and the project can be powered from a battery or a transformer. Short form kit only - you'll need to add your own power supply and enclosure.
- 9-12VDC at 300mA
- PCB and components

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BUILD THE SHAZAM!!

BY RON NEWTON

It was a dark and stormy night. Two small children were walking down the dimly lit sidewalk. The full harvest moon was peeking out every now and then through the small breaks in the clouds. A dog was howling on an eerie note. Spooks, goblins, and vampires could be seen on the other side of the street. The children slowly walked up to door with bags in their small hands. As they crept up closer, on the wall there was a strange picture of a skull or was it a lady sitting at her mirror? As they drew nearer, all of a sudden there is a great flash of light and Shazam! A lightning bolt from the picture flashes with a great clap of thunder!

Halloween is one of America’s (and my) favorite holidays. It has become progressively more and more popular with parties, and probably is in second for outdoor displays next to Christmas.

This project was designed to be flexible and reusable for other holidays, and the basic unit can be put together for less than $15 plus the board. No surface soldering is required, so this is an easy first time project. Depending on what type of display you use, the price can increase. Plain LEDs are cheap. I used high output white LEDs which sell for $.89 each. If you want 100 lumens per LED, the price can jump to $8 ea, and you will need to add drivers. For those who want lighting for a stage production, simply can use photo triacs (Sony S211602F) and drive up to 16 amps 220 volts for each bulb.

If you don’t want sound, leave out the sound chip and save money. The board can be triggered using a momentary switch. The one described in this article uses a Parallax Passive Infrared detector (PIR) for detecting the presence of a person. The PIR is also available at RadioShack for $9.99. My project ended up costing about $40. This doesn’t have to be a one-time project. I have put in programming pads, so if you have a PIC 2 or programmer you can program the light output for Christmas, Valentine’s Day, birthdays, etc. Changing the DIP switch will allow you to code many different programs. The voice chip can be reprogrammed by the DIP switch giving you many choices of different tunes. If you use the PIR, it can be utilized in a lot of other projects as it is a plug-in module.

Display Chip

The heart of the board is a Microchip PIC16F690. It also controls the sound module. The board will provide 30 milliamps for each of the 16 LEDs using a multiplexing technique. The eight outputs are divided into two sections of four outputs ($4^2 = 16$). The LEDs are set up in four groups of four with their anodes tied together. The first cathode of each group is daisychained together. Same with the second, third, and fourth LED. By applying a positive voltage to group one and grounding LED one, it will light; grounding LED two, it will light; and so on.

The PIC is triggered by shorting the two terminals labeled $S_2$. There is also a jack to plug in the PIR which will detect a person’s presence using infrared sensing. Both $S_2$ and the PIR provide a positive voltage to trigger the PIC. I added a three-position DIP switch so that you can program nine different modes ($3^2 = 9$). The switch also controls if the voice chip is being used for recording or playing. If you’re not using the PIR and if $S_2$ is kept shorted (using a shorting bar), the unit can cycle at 30 seconds, one minute, two minute, and four minute cycles,
Voice Chip

The voice chip is a CMOS device which uses five volts to power it. It has 10 seconds of memory for storing the sounds, a microphone preamplifier, and speaker outputs. I added a 1/8″ jack so that you can plug it into a stereo amplifier for big claps of thunder. Instead of using pushbutton switches as shown in the ISD1110P schematic which can be downloaded from the Nuts & Volts website (www.nutsvolts.com) or from Jameco.com. It is driven by the microprocessor so that the timing coincides with the lightning flash.

I got the thunder sound off of the Internet. The “PLAY” input was used as this allows the PIC to return to its high state and the voice chip will continue to run. The chip is capable of providing 12.2 milliwatts. Once it finishes, it goes to sleep.

Construction

The board files, along with several other downloads,
pads, and the three-position female headers in the PIR area. If using a three D, three C, or three AA battery pack, thread the wire through the strain relief hole and solder the red lead to the + and the black to the -. (All chips will run on 4.5 volts.) You can also use a five volt battery eliminator.  

**Do not use four batteries as the PIR will lock up!**

If programming with a PIC 2, make sure you place the DIP switches in the open position. The chip can also be programmed by other programmers if it’s removed from the board. The assembly files are also on the N&V website.

**Display**

The display I built is rather unique. I made a transparency of the picture drawn by Charles Allan Gilbert called “All is Vanity.” It is a picture of a woman sitting in front of her mirror and vanity. However, when viewed farther back, it becomes a skull. The picture is available on the N&V website if you want to use it.

The frame I used was the cheapest 8” x 11” stand-up frame I could find. The back was removed. I tacked the transparency on each corner to the glass using super glue. A sheet of white tissue paper was tacked to the transparency on each corner giving a semi transparent background but still showing off the picture. I took the back and added a 3/8” x 1” pine strip to make it look like a shadow box mount and painted the outside with a flat black paint. The inside was left unpainted. Holes were drilled in a zig-zag pattern with a #60 drill for the LEDs. The Superbrite LEDs were pushed through the back and held with wire wrap. The circuit board was mounted on the inside of the box in the left lower corner using 1/4” standoffs and 6-32 screws. However, I found that the PIR would not detect through either the transparency or the glass. I ended up drilling a 7/8” hole through the bottom of the picture frame and inserted the PIR into this hole from the inside, then tacked it with hot glue. I used a 3-pin male header and plugged it into the PIR female header, then wire-wrapped the PIR using the wire as an extension.

There are mounting holes in the board, and the 1/2”
6-32 screws will self tap. When the PIC is triggered, a bright flash of lightning comes from the picture. The frame was placed above my stereo speakers and the output from the voice chip was fed into the stereo input. A clap of thunder comes from the stereo speakers and really makes people jump.

If you don't want to use this particular picture, you can also mount the LEDs on a piece of white (how about a ghost hanging from the ceiling) or black felt (in a dark corner), depending on the effect you want. Zig-zag the LEDs in groups of four. There are a hundred and one possibilities!

Starting from the top of the zig-zag, use a wire-wrapping tool and 30 gauge wire wrap (available from RadioShack). Wrap the anodes of the LEDs in sets of four, e.g., LEDs 1, 2, 3 and 4 to pin one. Then do 5, 6, 7, and 8 to pin 2; 9, 10, 11, and 12 to pin 3; and finally tie 13, 14, 15, and 16 to pin 4.

Daisy chain the first LED's cathode of each set together and wire to pin 5, e.g., LEDs 1, 5, 9, and 13 to pin 5; LEDs 2, 6, 10, and 14 to pin 6; LEDs 3, 7, 11, and 15 to pin 7; and the last LEDs 4, 8, 12, and 16 to pin 7. It is important to keep the sets in order for the flash to appear going downwards (see Figure 7).

**PIR**

The PIR sensor is a pyroelectric device that detects motion by measuring change in the infrared (heat) levels emitted by surrounding objects. It uses a crystalline material that generates an electric charge when exposed to infrared radiation. If used, it takes about 60 seconds before it functions properly. The jumper should be placed on L. The PIR specifications state the voltage for the chip should be from three to five volts. Although it would fire, it would not activate the PIC using six volts.

**Software**

I placed the reading for the trigger at the beginning of the program so that you can change the DIP switch any time. After it is triggered, it will read the DIP switches.

Some of you who are new to programming may wonder what the "ANDWF" command does in the "STARTING AREA." The DIP switches for programming the chip are connected to port A's RA0, RA1, and RA2. These pins are configured to have pulled-up resistors, so all we have to do is to change the DIP switches from open to closed to configure the pins either high or low. There are other connections to port A besides the DIP switch. To read just the first three pins, you use an AND command. You can AND B'00000111' (B is for binary) or you could also use a "7" decimal number. A zero will eliminate any ones. The ones allow either a one or zero to be read. If we just AND port
### Parts List

<table>
<thead>
<tr>
<th>ITEM</th>
<th>DESCRIPTION</th>
<th>QTY</th>
<th>SOURCE</th>
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<tbody>
<tr>
<td>C1</td>
<td>220 μF 6.3 V</td>
<td>1 ea</td>
<td>Microchip.com</td>
</tr>
<tr>
<td>C2</td>
<td>.001 μF</td>
<td>1 ea</td>
<td>Jameco.com</td>
</tr>
<tr>
<td>C3 – C5 – C7</td>
<td>.1 μF</td>
<td>3 ea</td>
<td></td>
</tr>
<tr>
<td>C6</td>
<td>4.7 μF 16 V</td>
<td>1 ea</td>
<td></td>
</tr>
<tr>
<td>Header female</td>
<td>3 position .1”</td>
<td>1 ea</td>
<td></td>
</tr>
<tr>
<td>Header male</td>
<td>2 position .1”</td>
<td>1 ea</td>
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<tr>
<td>Header male</td>
<td>6 position .1”</td>
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<td></td>
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<td>Header male</td>
<td>8 position .1”</td>
<td>1 ea</td>
<td></td>
</tr>
<tr>
<td>IC1</td>
<td>PIC16F690</td>
<td>1 ea</td>
<td></td>
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<tr>
<td>IC2</td>
<td>ISD1110P</td>
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<td></td>
</tr>
<tr>
<td>Jack</td>
<td>1/8” mono</td>
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<td></td>
</tr>
<tr>
<td>LED</td>
<td>3 mm red</td>
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<td></td>
</tr>
<tr>
<td>R1-R6</td>
<td>1K 1/6W</td>
<td>2 ea</td>
<td></td>
</tr>
<tr>
<td>R2</td>
<td>5.1K 1/6W</td>
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<td></td>
</tr>
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<td>10K 1/6W</td>
<td>4 ea</td>
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<tr>
<td>R5</td>
<td>470K 1/6W</td>
<td>1 ea</td>
<td></td>
</tr>
<tr>
<td>R7-R8-R9-R10</td>
<td>150 1/6W</td>
<td>4 ea</td>
<td></td>
</tr>
<tr>
<td>S1</td>
<td>Three-position DIP</td>
<td>1 ea</td>
<td></td>
</tr>
<tr>
<td>*R12 jumper if not using PIR</td>
<td></td>
<td></td>
<td></td>
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<tr>
<td><strong>LEDs 1-16</strong></td>
<td>White 18000</td>
<td>16 ea</td>
<td>SuperBrightLeds.com</td>
</tr>
<tr>
<td>Motion detector</td>
<td>PIR Sensor</td>
<td>1 ea</td>
<td>Parallax.com</td>
</tr>
<tr>
<td>Wire-wrap wire</td>
<td>30 gauge</td>
<td></td>
<td></td>
</tr>
<tr>
<td>Battery holder</td>
<td>C or D (four battery)</td>
<td>2 ea</td>
<td></td>
</tr>
</tbody>
</table>

A, it will change its outputs which you don't want it to do. So, you place port A into a TEMP register and AND the TEMP registers instead.

After "START,” the strobe effect is created by using a series of flags and setting the flags after an LED is turned on. The first time it is turned on, it jumps back to S1. However, once turned on, its flag is set and it will continue on to the next LED. Note that all the flags are cleared at the beginning of the "START” sequence.

Once the strobing is finished, it turns on the voice chip and continues with a series of delays and flashes. The "CALL" feature is a great way of performing the same function over and over. You simply call the procedure you want; at the end of the procedure, you place a return and it will go back to the command after the CALL.”

### Running Shazam

Use either a small 16 ohm speaker or (better yet) plug your stereo into the jack. Set all the DIP switches to open. Put in the batteries. Place R12 across the PIR pins 1 and 3; short S2. The LEDs should flash downward and then the voice chip will activate. The LEDs will flash again. If you leave S2 shorted, it will flash every 30 seconds. You can change the DIP switches to increase the time. If you short and release, it will flash each time you short after the sound finishes at the time level the DIP switches have been set to. R12 prevents false triggering when using a switch.

Remove R12 and plug the PIR into the female headers on the Shazam board (if you are using it). When it detects the presence of a person, it will flash and wait for 30 seconds before flashing again. Delays can be set via the DIP switches so it won’t keep flashing.

### Recording

Remove the PIR module. Connect a microphone or tie the headers into the earphone outlet of your computer and connect it to the headers titled “MIC.” Put one DIP switch in the closed position and the other two in the open position. When ready, short S2, open S2, and short again to record. The LED will light indicating that the chip is recording. Wait until the light turns off.

There are many thunder sounds available on the Internet. Just perform a search for “thunder sounds.” Use a 1/8” plug with about 6” of wire-wrap; put the plug into the earphone jack of your computer and wrap the wire around the “Mic” pins. Push both the play button on your computer and short S2 at the same time to record. The chip will record 10 seconds of sound. If you goof, try again.

Happy spooking!! NV
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MAKE MAGIC CANDLES WITH THE PROPELLER

If you’re like me, October is a very busy month, mostly trying to finish building props and decorations (some from last year!) for Halloween. Last minute ideas abound, but most just don’t fall into the time constraints I usually find myself boxed in by.

A really great way to add atmosphere to a Halloween display is with candles. The real ones, though, can be dangerous, go out if the wind blows, and cannot be controlled with any sort of [practical] automation. There are electronic candles available in retail and specialty stores, but most are not terribly bright and — like their real counterparts — cannot be controlled (without some major hacking).

There is good news. With some simple parts, you can whip up a custom candle controller that will run eight outputs, and can easily be expanded for more. In addition to candle outputs, there is a trigger input to control behavior, and at the end I’ll show you how a few more components can be added for additional control; for example, playing audio from an external device.

Building a Candle Controller

My processor of choice is the Parallax Propeller chip. Having worked with it almost daily for the past 18 months, I have a real comfort level and enjoy developing code for it. It’s high level language, Spin, is fairly easy to use and the interactive nature of the IDE simplifies and expedites experimenting with new code.

In order to make this an “afternoon project,” I hand-wired the circuit to a Propeller proto board. These boards are an inexpensive way to get started with the Propeller. If you’re going to do this as a one-off and never use the Propeller again, you can get a USB version of the proto board. If you think the Propeller is in your future, then you can go with the cheaper (non-USB) version and use a Prop-Clip to program it.

Let’s look at the circuitry first. The candle outputs are driven by a ULN2803A; this lets us use anything from simple LEDs to some medium current monsters. With eight outputs running, you should limit the current to about 180 mA through each ULN channel. When it comes to LEDs, that’s a LOT of current. Figure 1 shows the ULN connections to the Propeller. Advanced users will note that resistors R1-R8 are, in fact, optional but having seen ULNs melt internally (usually from an oversized inductive load) do bad things on the processor side, these resistors are cheap insurance.

There’s nothing to the “wick” circuit shown in Figure 2. I’m using a 12 volt supply and the resistor shown allows about 20 mA through the LED. The LED, in fact, is the trickiest part of the project. Finding one with the right color and brightness can be a chore. I found that the NTE30039 has the right color (nice warm orange) and is very bright (7,000 mcd).

We could stop there and have a lot of fun, but why stop? On Halloween, the ToTs (Trick-or-Treaters) will wander
up the sidewalk and it might be fun to have the candles change as they approach. The trigger circuit shown in Figure 3 accommodates a Parallax PIR sensor (via a three-pin header X1) or a normally open contact (usually a mat switch) connected to TB1. The voltage for the PIR and the normally open contact is 5V, so a 2.2K series resistor is used to limit the current into the Propeller (which operates at 3.3V).

Construction is point-to-point and as it can be difficult to see these connections in a photo, I created the graphic in Figure 4 as a guide. Note that the resistors between P0-P7 and the ULN2803A are stood up on end. Also note that while it looks like the wires go on top, they are actually soldered to the bottom. Finally, the colored dots on the PIR header show cable orientation (ground is at the top of the board). Figure 5 shows my completed prototype.

With a few items from the hardware store, we can turn the simple LED circuit into a nice looking flame. The first thing we have to do is modify the LED. I’ve never been able to find bright LEDs that are diffused; the bright ones tend to come in a water-clear package. This makes the output very beam-like, as the convex end of the LED acts as a lens. We can fix that with a file or a bit of sandpaper. By grinding off the dome at the end of the LED, we will get more light shining from the side of the package. Figure 6 shows unmodified and modified LEDs.

The next step is to solder the resistor to the LED, add leads, and then protect everything with heat shrink tubing. In addition to protecting the connections, the heat shrink adds a little stiffness that will be useful in the final step. Figure 7 shows a wick in process and one that is completed. I tend to solder the resistor to the anode side of the LED.

If you want to create a flameless faux candle, you can use the wick as is; simply install it in a candle body, recessed enough to provide the illumination created by a hidden flame. You can actually take real candles (pillar type), burn them down a bit, and then drill a hole through the center of the body. By installing the LED in the wax candle, the light will shine through the translucent paraffin and create a beautiful effect.

To create a “flame” for the wick, I use clear silicone—the type used to seal seams around bathtubs and sinks. The problem with silicone is that it takes an annoyingly long time to fully cure. This can be fixed by adding two drops of water and two drops of glycerin into two ounces of silicone. In order to ensure a complete mix, I dip the end of a match stick into yellow acrylic paint and then dab that into the silicone. (A tiny bit of color goes a long way.) The idea is not to color the silicone so much to change it, but to provide a visual indication that the water and glycerin are completely mixed in. By adding the water and glycerin, the wicks will be ready in a couple hours versus the 24 normally required for the silicone to cure.

After mixing the silicone in a small cup, dip the LED into it, swirl it around a bit, and then slowly pull it out to create a natural, wispy flame shape. You must keep the LED in the upside-down position until the silicone is completely cured (give it two hours if you mixed in the water and glycerin). Figure 8 shows a completed wick/flame.

For many projects, I build faux candles from PVC pipe. Simply cut it, clean it, and apply hot glue drips to the top to finish the body. The wick is installed and secured with hot glue; at this point, I have a candle ready for a
prop or set. **Figure 9** shows a small candle created with my LED wick and a piece of 1/2” PVC pipe.

### Programming a Jack-o-Lantern

Even if you’re fairly new to using Spin, you should find my demo program easy to modify after I’ve run you through the explanation. For those that are using the Propeller for the first time, I think you’ll find its behavior and programming interesting.

The program is composed of three elements: 1) the main code; 2) a dimmer module to control LED brightness; and 3) a flame effect generator.

Each of these modules runs in their own cog (processor) inside the Propeller so we don’t have to worry about managing interrupts (the Propeller doesn’t have them) or other time-slicing strategies; trust me, multiple processors on one slab of silicon makes life very pleasant.

When the Propeller boots up, it will load the Spin interpreter into Cog 0 and run our main program. Inside that program, we will instantiate the other cogs. The first is a dimmer cog that runs jm_bam8; this is used to modulate the eight LED outputs using a process called Bit Angle Modulation (for details on how BAM works in the Propeller, see The Spin Zone column in the November ’09 issue of Nuts & Volts). We don’t have to understand how BAM works to use it, just what to do with the object.

The first thing we need to do is include it in the code. This is accomplished by adding it to the objects list in the main program:

```c
obj leds : "jm_bam8"
```

This includes the file jm_bam8.spin in the compilation and creates an object called **leds**. Most objects that deal with I/O will require some sort of setup; usually with a method called **start()** or **init()** — the latter is what I tend to use in my objects.

To instantiate **leds**, we’ll call its **init()** method with the first pin of an eight-pin group; in my program, there is a constant called **WICK1** that is that pin (0 for the hardware we just built):

```c
leds.init(WICK1)
```

This line of code launches the BAM processor into Cog 1 and will continue to run until we stop it with the **finalize()** method. To change the brightness of any of the LED wick outputs, we can use the **set()** method with a value from 0 (off) to 255 (fully bright). Having just discussed the BAM object and knowing that our program needs a flame generator, you may be wondering why that object is not in the object list above. Let me show you. Here’s the main loop that runs my Jack-o-Lantern:

```c
pub main | ch, bright, delay

lottery := cnt

leds.init(WICK1)

bytefill(@level, 0, 8)

cognew(flicker(3), @stack)

repeat ch from 0 to 2

  repeat bright from 0 to 255

    pause(4)

    level[ch] := leds.ezlog(bright)

  pause(2000)

repeat

  repeat bright from 255 to 64

    pause(4)

    bytefill(@level, leds.ezlog(bright), 3)

    flickerdelay := (255-bright) * 52 / 10

    waitstart(150)

  ?lottery

  delay := (11*lottery//8001) + 2000

  pause(delay)

  flickerdelay := 0

  repeat ch from 0 to 2

    level[ch] := 255

    pause(750)

    pause(30000)
```

---

### PARTS LIST

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<th>ITEM</th>
<th>DESCRIPTION</th>
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<td>2N3904, NPN</td>
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<td>330 ohm</td>
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<td>R17</td>
<td>2.2K ohm</td>
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<td>10K ohm</td>
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<td>S-S1</td>
<td>Socket, 18-pin</td>
<td>Mouser 571-2828362</td>
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<td>Term block</td>
<td>Mouser 511-ULN2803A</td>
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<td>JMPR</td>
<td>0.1 shunt</td>
<td>Mouser 517-350-00</td>
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<td></td>
<td>Parallax 32212</td>
</tr>
<tr>
<td></td>
<td></td>
<td>Parallax 32812 (USB)</td>
</tr>
</tbody>
</table>

Hook-up wire
Heat shrink tubing
Clear (bathroom) silicone
Water
Glycerin
Acrylic paint (yellow or orange)
In the course of the program, I want to do some random timing so I've created a variable called \texttt{lottery} to hold a random number. This variable is initially seeded with the value of the system counter. I like to do this because the Propeller boots using an internal RC clock source which will always have a bit of variance.

We’ve already discussed starting the BAM driver, so we can move on to getting the flame simulation going. An eight-byte array called \texttt{level} will be used to control the maximum brightness of each candle output. As with the BAM driver, the values in \texttt{level} are 0 (off) to 255 (maximum brightness).

All of the output levels are set to zero by using \texttt{bytefill}. The next step is to start the flame generator which is, in fact, a method called \texttt{flicker()} that is part of the main code listing. Yes, that’s right: We can launch a method into its own cog to run parallel with other processes. The cool thing about using this technique is that the method we launch can still access global variables and other methods in the program.

This is how we launch the \texttt{flicker()} method into its own cog:

\begin{verbatim}
cognew(flicker(3), @stack)
\end{verbatim}

In this particular case, we’re telling the flicker generator that we have three wicks (pins 0, 1, and 2); by setting the actual number of wicks used, we have more flexibility in how fast the routine runs.

When using \texttt{cognew} to launch another Spin cog, we need to provide the name of the method (along with parameters, if any) and the address of a block of longs that will serve as stack space. The stack is used to hold local variables, as well as working variables and return addresses so that other methods can be called.

To be honest, setting the stack size can be something akin to black magic — this is not unique to the Propeller. In the past, I’ve tended to start with 16 and then add up the number of local variables used in the method. That strategy didn’t work in this case and the program just stopped. As soon as I bumped up the stack, everything worked fine. I generally believe it’s best to be generous with stack space.

If you get concerned about memory usage, you can find tools in the Parallax Object Exchange that will analyze stack usage and let you optimize for the smallest RAM footprint.

Now that all of the elements are running, we can get into the working code. I start by ramping up the brightness of each candle output just so I can see that they’re working. With 256 steps and a 4 ms delay between each, it will take about a second for each candle to light to full brightness. This holds for two seconds and then drops into the loop that will run until we power-down.

I often joke that my Halloween prop control programs are like a Quentin Tarantino movie in that they start at the end and then loop back. You can see this at the start of the loop: The brightness of all candles is being dimmed (from 255 to 64) and the flicker speed is being modified to go from 0 (fast, angry) to 1,000 (slow, calm). At the end of this loop, we’re back at the idle state with calm, low brightness wicks.

You’ll remember that the circuit is set up to accept a Parallax PIR or a simple, normally open switch (usually a mat switch). We’re going to use this input to bring Jack to life when a ToT approaches. It’s always a good idea to debounce digital inputs and it is especially important with PIRs as they can be twitchy. We can wait for a good input with a call to \texttt{waitstart()}:

\begin{verbatim}
waitstart(150)
\end{verbatim}

This call waits for the start input to be active — and stay active — for 150 milliseconds before being considered a valid input. This timing will prevent spurious outputs from the PIR from triggering the prop. The \texttt{waitstart()} method is pretty straightforward:

\begin{verbatim}
pub waitstart(ms) | debounce
ms := 10 >= ms <- 1_000
debounce := 0
repeat until (debounce => ms)
  pause(1)
  debounce := ++debounce + ina[START]
\end{verbatim}

This method “fixes” the input parameter so that the valid range for debouncing an input is 10 milliseconds to a full second. Spin has a numeric range from negative two billion (plus) to positive two billion (plus) — this little correction keeps a typo from hanging up the program.

Here’s how the method works: A counter (called \texttt{debounce}) is cleared and then we drop into a loop that waits one millisecond before incrementing \texttt{debounce} and then multiplying it by the state of the START input pin. If the pin is active (1), then the count will be maintained; if the input is not active (0), then the \texttt{debounce} variable will be cleared. This process ensures that the input is active and stays active for the desired period. Once that happens, we return back to the main program.

In my props, I like to insert a randomized delay between the trigger event and the active prop code. To do this we randomize the value in lottery with \texttt{?} and then perform a little math to get a value between two and 10 seconds. Note that we have to take the absolute (\texttt{| |}) value of \texttt{lottery} in the code; this is necessary because a negative value (which is possible with the randomization) will create a problem for the modulus (\texttt{/}) operator.

After the random delay, the flicker rate is set to zero which makes the flames more active and then each is bumped to full brightness with a short delay in between. It’s easy and very effective.

Here’s why: Imagine ToTs approaching your home where they see a friendly Jack-o-Lantern on the porch, gently lit and looking peaceful. As they draw near, the flames are magically brightened and seem a bit angry. In the minds of the ToTs, you’ve just created a “What’s next?” scenario. They will be [happily] on edge until they’re safely back on the sidewalk with the treats you’ve just provided. The reason for the random delay after the trigger point is to fool returning ToTs.

The candles burn brightly for 30 seconds before going back to the top of the loop (where we started) and are
dimmed back down to the idle level.

Again, this is very simple and with a little extra hardware
and code you could even add audio control. The circuit in
Figure 10 can be used to start audio players that require an
open collector start pulse, e.g., Cowloquus (set the jumper
to select the transistor and pulse the TX pin low-high-low),
or to communicate with a serial audio player (e.g., EPX-TEK,
Rogue Robotics) when the transistor is bypassed. Serial devices
will need a serial code but that’s not a problem as the base
Jack-o-Lantern code only uses three of the eight available.

I suspect many will stop right here and head to the
workbench to build a candle controller. For those hearty types
who want to know how I generate realistic flames, read on!

**Mimicking Nature is Difficult**

My friend, Jen, was having dinner at a nice restaurant
and noticed they were using electronic candles — and they
looked awful. I understand; having experimented with
candle algorithms, I can tell you it’s a little tricky to get a
natural look (though it won’t be after you read this!).

There is good news: We don’t have to do a perfect
simulation. You see, our brains have this incredible ability
to take partial information and fill it out with what we
know; this gives us the ability to recognize friends from
bad photos at odd angles. So, you see, we only have to
get close; our brain will say, “Okay, that’s a candle.”

Don’t believe me? A few years back I had two candles —
one real, one electronic — “burning” side by side so I could
watch them. The air conditioner kicked on and blew out the
real candle. Without thinking, I picked up a match and held
it to the electric candle expecting it to light. After about a
second, I laughed at myself — out loud! — for being fooled
by my own creation. This illustrates the power of our subconscious
and its ability to deal with partial (even false) information.

After lots of experiments, I’ve developed an algorithm
that works like this: Each “wick” requires three variables:
1) current brightness; 2) target brightness; and 3) ramp rate
moving from the current brightness to the target. When
the current level reaches the target level, a new random
target is generated. Of late, I am also keeping track of the
directional change to ensure that the next update always
passes through the mid-point of the brightness range.

**Figure 11** is a simplified illustration of the algorithm’s
behavior (the ramp is actually a bit “wobbly” because the
ramp variable is reseeded when it hits zero). When the
new direction is ramping up, the brightness will be selected
in the upper region; when the new direction is ramping
down, the brightness is selected in the lower region. This
seems to give the wicks a bit more life (as would want
in a display). In the final code, I also apply an overall level
control so I can control the maximum brightness.

**Coding Candles**

Keeping track of up to eight “live” wicks really dictates that it happens as a background process — this allows us
to monitor sensors and other things in the main code
without affecting wick behavior (unless we want to). As
discussed earlier, the process is coded into a Spin method
which launched into its own cog:

```plaintext
pri flicker(wix) | lotto, idx, tmp, direction { 
  base[8], target[8], ramp[8] 
  
  lotto := cnt
  repeat idx from 0 to 7
    target[idx] := ?lotto & $F0 | $80
    base[idx] := 0
    ramp[idx] := (?lotto & $1) + 1
    direction := %00000000
  repeat 'flicker loop
    ?lotto
    repeat idx from 0 to (wix-1)
      if (!ramp[idx] == 0)
        ramp[idx] := (?lottery & $11) + 1
      if (base[idx] < target[idx])
        base[idx] += direction
      elseif (base[idx] > target[idx])
        base[idx] -= direction
      else
        direction := togglebit(direction, idx)
      
      if (getbit(direction, idx) == 0)
        target[idx] := ?lotto & $F0 | $80
      elseif (target[idx] > ?lotto & $7F)
        target[idx] := ?lotto & $7F
      
      tmp := (base[idx] * level[idx]) / 255
      leds.set(idx, tmp)
    
    if (flickerdelay > 0)
      tmp := (flickerdelay * US_001) #> 400
      waitcnt(tmp + cnt)
}
```

Okay, it looks a little scary, right? Hey, this is the
Halloween issue! But just like your friend behind the mask,
once you see what’s underneath you’ll see that it is not as scary as it seems at first blush.

There are two loops: the first sets up initial values; the second runs the active flicker algorithm. Most of the variables used by the `flicker()` method are local, but you’ll see that it uses the external `level` array and the `leds` object to control the wicks without further intervention by us. The code also accesses the (global) `flickerdelay` variable which affects the speed of wick movement.

At the top, a local random variable (lotto) is initialized and a loop runs which sets the base value of each wick to zero — the target in the upper range — and then randomizes the ramp control. The `direction` bits are cleared to zero which indicates that each wick is initially ramping up.

Now for the fun. At the start of each loop iteration, the value in `lotto` is re-randomized so that it’s always changing. For each wick, we decrement the ramp value and when that hits zero we increment or decrement the base value so that it moves toward the target. Once the target is reached, a new (random) target is generated.

Whenever a wick brightness is modified, the new level is set in the `leds` object. Note the use of the global `level` array which acts as an overall brightness control value. As we modify the `level` array in the main cog, it is used by the flame cog. I think this is really cool.

Overall loop speed is dictated by the number of active wicks and the value in `flickerdelay`. As we saw in the main code, we can change `flickerdelay` to alter the apparent behavior of the wicks. You’ll see there is just a bit of math before the `waitcnt` command — this is needed to prevent `waitcnt` from hanging (delay of about 56 seconds) when we have a small value in `flickerdelay`.

This process could be duplicated in another micro but I believe it would be a much bigger coding challenge.

---

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Another great thing about the Propeller is that you can very quickly test new code by downloading straight to its RAM (all programs run from RAM). This lets us make and test changes very easily. We can do this by pressing the F10 key in the IDE. Once everything is the way we like it, we can do a permanent download of the final code with F11.

Well, there you have it; a fun Halloween prop and a neat way to safely simulate flames. Want a hellish pit of burning coals in your Halloween display? No problem! Just use red and orange LEDs and set the flicker algorithm to low-and-slow. With a little imagination and a bit of fine-tuning of the flicker parameters, you can simulate most anything that’s “burning.” Happy Halloween! NV

Jon McPhailen is an actor, writer, and electronics enthusiast based in Los Angeles, CA. He portrayed John F. Kennedy in several episodes of The History Channel’s “Unsolved History” series and can be seen with Tom Wilkinson and Frances Fisher in the hit comedy, “The Night of the White Pants.”
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IMPLEMENTING A FILE I/O SYSTEM FOR THE 16-BIT MICRO EXPERIMENTER

In a previous Nuts & Volts article, we showed a Graphics Module add-on for the 16-bit Micro Experimenter (Experimenter for short). In this article, we will discuss adding more capability to our Experimenter by integrating a PC compatible file I/O. The SD card technology and Microchip’s freely available file I/O library, along with the Parallax SD carrier board makes this all possible. The software and hardware footprint is fairly minimal making the file I/O system an easy add-on module for your Experimenter applications. Gone are the days when you have to labor with custom solutions to move data between your PC and microcontroller. Also gone are the days when you are constrained in the amount of data you can transfer. With this new capability, you essentially have a low cost, small form factor multi-GB removable “hard drive” that can work seamlessly with either a PC or the Experimenter. It will allow both devices to independently create, write to, and read from, delete, and rename data files. This article will discuss the Microchip library briefly, how it was modified for the Experimenter, and the specific hardware and software needed for the Experimenter. It will also cover several demo applications. In this, as well as all other previous articles on the Experimenter, software is written in ‘C’ so some knowledge of its basic syntax will be required to understand the file I/O library and its API.

Figure 1 shows the Experimenter connected to a 2 GB SD card through the Experimenter I/O expansion port running one of the demo applications (DEMOREADWRITE). It uses an SD carrier board offered by Parallax.com for easy insertion and removal of the SD card within the Experimenter’s solderless breadboard environment.

The SD Card

Let’s review the SD card. It is difficult these days not to notice Secure Digital (SD) cards (see Figure 2). They appear just about everywhere. An SD card is a non-volatile Flash memory card format developed for use in portable devices. It is widely used in digital cameras, digital camcorders, and portable laptop computers to name a few. Its size and the format factor make it ideal for adaptability of use with microcontrollers like the Experimenter.

The format has proven to be very popular. Standard SD card capacities are up to and exceed
2 GB. These cards ship preformatted with a FAT16 or FAT32 file system. This is the exact file format for all of Windows' machines (we will discuss this more a little later). Because of this, the SD card can be accessed on virtually any PC with an SD reader. Once inserted into a PC reader, Windows allows the SD card to assume the "identity" of another disk drive within the PC. At that point, Windows' operating system can be used seamlessly with the SD card files, and file data can be read as long as Windows software recognizes the file extension associated with the file. For these demos, I restricted myself to Secure Digital SD cards using card sizes of 256 MB, 512 MB, and 2.0 GB. They are readily available, and my suggestion is to use one for these experiments.

**Background on FAT16 and FAT32 File Systems**

Just to give some background on this subject, in order to store files, the PC needs a filing system that defines the names of the files, as well as a system to track which sector (512 bytes or smallest traceable data "chunk" in a file system) is stored in which file. In addition to reading and writing data, the system should be able to create, delete, and rename files. The File Allocation Table (or FAT) is used as the filing system or "personal card catalog" that supports these functions for both the PC and the SD card.

The job of FAT is to keep track of all files on a given system. FAT16 is the oldest of the Windows file systems and the 16 refers to the fact that this file system allocates 16 bits to store addresses of each sector. If this number is too small, it may be necessary to upgrade to FAT32, on the other hand, uses 32 bits as a sector address to support up to 2 TERABYTES of data.

The Microchip library is configured to support both formats. This provides a lot of removable data storage capability for microcontrollers (like the Experimenter) to take advantage of.

**The SD Card Hardware Interface**

The hardware interfaces are shown in Figure 3. The card interface itself is shown first; then the card with the carrier interface second; and finally a block diagram of the entire system connected to the

Experimenter. Note that SD cards operate from 2.0V to 3.6V and this is easily accommodated with the Experimenter +3.3V operations. SD cards support nine electrical contacts in addition to two contacts for insertion detection and write-protection switch setting. The nine electrical contacts can work in two distinct modes of communication: the native original SD bus (4-bit parallel interface); or an SPI (Synchronous Serial Interface) mode. SPI is the mode we will use for the Experimenter. When using SPI, not all of the electrical pins are used and are therefore tied high. For the Experimenter, that leaves us with just eight signals in total that we must support (see Figure 4).

- **CD** — SD card detect
- **CS** — Card select
- **WP** — Card write protection setting
- **DO** — SPI serial data out of card
- **DI** — SPI serial data into card
- **CLK** — SPI clock for data transfer (originates in Experimenter)
- **Power, GND** — +3.3 VDC and ground from the Experimenter

The PIC24F microcontroller on the Experimenter has two internal SPI peripherals, SPI1 and SPI2. All the demos configure the Experimenter I/O expansion bus for SPI2 and digital I/O. The SPI2 supplies three basic pins for communication: Serial Data in (SDI), Serial Data out (SDO), and Serial Clock (SCK). Chip Select (CS), card detect, and write-protect are configured as digital I/O. The final I/O expansion bus

![Image](image-url)
interface tally is as follows:

- Pin 4 SPI2 Data Out — Connects to SD card DI.
- Pin 5 SPI2 Clock Out — Connects to SD card CLK.
- Pin 6 SPI2 Data In — Connects to SD card DO.
- PIN 2 Digital In — Connects to (WP) Write Protect (this is low when write-protect is off).
- Pin 3 Digital In — Connects to (CD) Card Detect (this is low when card is present).
- Pin 7 Digital Out — Connects to (CS) Card Select (this is low when card is selected).

A final hook-up diagram using the Parallax SD card adapter and the Experimenter is shown in Figure 5. Note that an activity LED is connected to the CS line of the SD card. When the LED is “on,” the SD card is actively being accessed from the Experimenter. When the LED is on, don’t remove it from the carrier; it is indicating that the Experimenter is actively accessing the card.

To use the SPI2 within the PIC24F, we need to configure the PPS or Programmable Peripheral Select capabilities of the PIC24F. PPS exists as a feature for the smaller pin packages of the Microchip PIC24F family. It allows users to configure where they want the internal peripherals to appear on the outside pins of the package. This is done for the simple reason that the internal peripheral count is too large to be accommodated by the limited pin count of the package. For the PIC24F, there are a total of 16 programmable pins on the package. Ten of these are available for use on the Experimenter’s I/O expansion port. The details of this PPS programming can be seen in the main code of each of the demo projects. Details of how PPS is done in general can be found in the datasheet available from Microchip on the PIC24FGA004 series, as well as in the beginner’s experiments lesson plans available at www.KibaCorp.com.

**Microchip File I/O Library**

Microchip developed a file I/O library that is freely available to all users of Microchip products. It is configurable and allows users to only compile and accept those library
elements that they want to use in their application. An excellent resource is the Microchip Application Note AN1045 “Implementing File I/O Functions Using Microchip’s Memory” that is available on their website at www.microchip.com. We adopted this library for the Experimenter, and the applicable API (Application Programming Interface) calls shown below:

- **FSInit** — Initializes card and interface.
- **FSiClose** — Updates file info and closes file.
- **FSfeof** — Verifies if the end of the file has been reached (used in read operation).
- **FSfopen** — Opens existing file or creates new one.
- **FSfread** — Reads opened file.
- **FSfseek** — Changes the file pointer position in file.
- **FSf tell** — Returns current pointer position in file.
- **FSfwrite** — Writes information to file.

There are two other functions we need to consider:

- **MDD_MediaDetect** — Determines if SD card is in carrier.
- **SetClockVars** — This API sets the timing variables manually that are used to set the file create/modify/access times. This function is used only if we elect not to use the internal Real Time Clock Calendar (RTCC) peripheral to set time.

Let’s go through a couple of examples. There are a total of three demos: simple file write WRITEDEMO; simple file read READDEMO; and a larger example that uses two files (READWRITEDEMO), the RTCC peripheral, UART, and LCD, as well.

**Configuring the Microchip File I/O Library for the Experimenter**

Before we proceed, let’s exam how the library was configured to run on the Experimenter. This section let’s us look “under the hood” so to speak. Microchip offers lots of free libraries and working through

- FIGURE 7. Library reference and hardware configuration in Main.

The hardware interface for the library is configured in two places. The first place is Hardwareprofile.h. The other place is in Main (). Here, we map SPI2 using PPS (described earlier) to the appropriate Experimenter I/O pins. Note that this example can help you understand what is involved in adopting these libraries for your own use. Keep in mind that all our demos are preconfigured and can be used “out of the box” as a template for your applications.

The library options are configured in FSconfig.h by selecting which macros you want to enable. For the Experimenter, the following is enabled:

```c
#define FS_MAX_FILES_OPEN 1
#define MEDIA_sector_SIZE 512
#define MEDIA_allow writes
#define MEDIA_allow FORMATS
#define SUPPORT_FAT16: supports both FAT16 and FAT32 file formats
#define USER_DEFINED_CLOCK: allows use of PIC24 RTCC (Real Time Clock, Calendar) to automatically set time stamp for file creation and update
```

- FIGURE 6. Library Hardwareprofile.h.

file I/O library reference

unlock for PPS

SPI2 PPS programming to pins

lock PPS

In MAIN () Setting up Library for Use with Experimenter

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there is an un-lock and lock sequence that surrounds this PPS programming to insure that this critical function is not invoked haphazardly. Also take note that only include FSIO.h is needed in Main() to reference the entire Microchip library.

**DEMOWRITE - Open a Text File and Write Data to It**

Our first example creates a file and writes the following line to it: “This files I/O test using Microchip library.” The LCD library is used to display the status. Download and unzip WRITEDEMO from the website. Open the folder and double-click on MDDFS_SD_PIC24.mcp. In Step 1, we call MDD_MediaDetect(). This function determines if the SD card is present. After Step 1, we then execute Step 2: FSIInit(). This function initializes the library and the SPI port. Step 3 opens a file “FILE1.TXT” with write or the “w” attribute. In this step, the FSfopen() will create the file if it doesn’t exist; if it does exist, it will clear it of previous data. Step 4 does the actual write of data to the file using FSfwrite(). This function takes data preloaded in the send buffer array and writes it to the file one byte at a time. In Step 5, we finally close the file with FSfclose(). After executing WRITEDEMO, we can remove the SD card from the Experimenter and place it in the PC SD card reader to verify the file contents. With the SD card in the PC, it will register the card as a hard drive.

Open up the hard drive and note that FILE1.TXT exists; double-click on the file (Windows Notepad should automatically open it) and the proper text should be displayed. It should read “This a file I/O test using Microchip library.”

We can create files with any extension that Windows can recognize. However, .TXT is the simplest. This format works well for data written in an ASCII format. If you write lots of data records to your file, keep in mind that you need to use some kind of delimiter between your data records like a space or comma.
DEMOREAD - Open a Text File and Read Data From It

Okay, so we were able to write using the Experimenter and read using the PC, but let's double-check that the Experimenter can read the file it created. Download and unzip READDEMO from the website. Open the folder and double-click on MDDFS_SD_PIC24.mcp. Again, the LCD library is used to display the status. Let’s remove the SD card from the PC and put it back in the Experimenter SD carrier board. Note the use of the SetClockVars() in the beginning of the code. Steps 1 and 2 are identical to the earlier write code. In Step 3, we open the file for reading using FSopen(), using the FILE1.TXT file name, and an “r” or read attribute. In Step 4, we do the actual reading with FSread(). The entire content from the file is read one byte at a time and loaded into a receiveBuffer. We then close the file in Step 5. In Step 6, the LCD display is used to show the entire contents in receiveBuffer. You can exam the LCD for a complete string readout for “This a file I/O test using Microchip library.”

DEMOREADWRITE - A Larger Example Using Two Files, File Pointers, RTCC, LCD, and Serial I/O

This demo creates two files (one at a time) and uses the RTCC to set the date/time of file usage. In addition, both the UART and LCD libraries are used. The UART is totally optional but does require an Acroname adapter (covered in NV June ‘10 issue) and setting HyperTerminal to 9600 8N1 (see the hook-up diagram shown earlier) to display the contents on your PC. Download and unzip READWRITEDEMO from the website. Open the folder and double-click on MDDFS_SD_PIC24.mcp.

This demo first writes a “This is Experimenter string!” to FILE1.TXT; it then copies this to FILE2.TXT and rewrites the last character in FILE1.TXT to a 2. The results are shown in Figure 14. All steps are displayed on the Experimenter LCD simultaneously as they’re being written to HyperTerminal.

Future Expansion

We’ve now added a pretty substantial capability to our Experimenter. Data logging and data sharing have reached new heights — we now have GB storage and retrieval capability in a microcontroller environment and the ability to use Windows-based tools seamlessly with the data. Think of the applications! NV
PHREAK OUT WITH THE \nBASIC STAMP 2

By L. Abraham Smith N3BAH

For many years, hackers built special electronic devices known as “blue boxes” for the purpose of transmitting special audio tones over telephone lines. The practice — known as blue boxing — provided challenges to the subcategory of hackers known as phone “phreaks.” They had to be able to both construct the tone oscillator circuits to produce the telephone company’s proprietary audio tones, and devise how to employ them to explore the telephone switching networks. While the practice of blue boxing died out in the 1990’s with the introduction of electronic switching systems which were not susceptible to in-band tone manipulation, interest in the technical aspects of telephone communications continued.

Now, a new generation of telephone experimenters are phreaking out over Asterisk™ (www.asterisk.org) — the open source Linux program which can put your own private telephone system inside your computer! The proliferation of computers running Asterisk interconnected over the Internet and interfaced to the public switched telephone network (PSTN) has given rise to a new form of legal phreaking where users configure their Asterisk servers with special features which can only be accessed by users transmitting the old-school blue box tones in appropriate combinations. This project which has gained worldwide popularity is known as Project MF (www.projectmf.com). Briefly stated, Project MF recreates the techno environment which existed “back in the day,” making it possible to experience the challenges of blue boxing from the legal safety of the Asterisk VOIP system.

Recently, I began experimenting with Asterisk and Project MF. I needed a low cost programmable audio tone generator capable of producing standard Dual-Tone Multi-Frequency (DTMF) telephone touchtones, as well as unique Single- and Multi-Frequency (MF) tones. I wanted to design a circuit using readily available components which was easy to assemble and which could be modified by firmware changes as my needs evolved. The solution I implemented uses a microcontroller and provides the novice builder with an inexpensive and fun way to learn about the interaction of micro hardware and software while constructing a unique and useful tone generation device.

Why a Microcontroller?

While there is no shortage of PC software which could generate the necessary tones, what fun is that? In order to capture the full effect of the retro-tech environment presented by Project MF, you need to build a hardware blue box. I found a number of schematics on the Internet, most of which either required obscure tone generator ICs or set the tones with tedious potentiometer-tuned circuits. I wanted a circuit which would use readily available parts and allow tones to be easily programmed and changed via embedded software. The desire for a combination of hardware and modifiable software in one package led me to consider a microcontroller.

Most microcontrollers share basic features with each manufacturer’s products having their own unique instructions or programming objects which make them better suited for certain tasks. My experience has demonstrated that the BASIC Stamp 2 from Parallax, Inc., is uniquely suited for the purpose of audio frequency generation. There are less expensive micro platforms than
the BS2, but I chose this module because it is strongly supported with free tutorials and programming documentation from the Parallax website and numerous online user groups. It programs in PBASIC which is easily learned by hobbyists who are not experienced in programming. After you’re finished experimenting with the project described here, you will have the confidence and skills to re-use the BS2 in other projects with programs you create!

With the BASIC Stamp, FREQOUT and DTMFOUT do all the work

The kit for the BASIC Stamp 2 OEM module (Parallax part number 27291) is easy to assemble even for the novice builder, and includes a PCB (printed circuit board) with all the components and their locations clearly marked. It is easily programmed via the on-board serial port in the PBASIC language developed by Parallax. One of the unique features of the BS2 is its ability to generate standard touchtones, as well as custom MF tones using the PBASIC commands DTMFOUT and FREQOUT which make it possible to generate both standard telephone Touchtones as well as any other single- or dual-frequency tones and sequences. This micro module is well supported with free reference materials and IDE (Integrated Development Environment), and serves as a good intro to microcontrollers and their practical uses (which will doubtless spur the user to devise project ideas of their own). Finally, the BASIC Stamp 2 is available from a number of sources.

Here you’ll learn how to interface the BS2 module with a keypad, a few switches, and a handful of discrete components, then load free open source firmware to make your own unlimited tone generator box. The concept is simple: The hardware checks the state of the 12 keys on the keypad in a continuous cycle, along with monitoring the state of the toggle and pushbutton switch; it then instructs the BS2 to generate the appropriate touchtone or custom tone(s) output via pin 0. By simply changing the values in the software and loading it onto the BS2 via the integral serial port, you can generate any single- and multi-frequency tone(s)/sequences without changing the hardware. Tone frequency, duration, pauses, and sequences can all be modified according to your needs. (More on the programming later.) Now, let’s build the PhreakEncoder!

The Build

Step 1: Assemble the BS2.

The BS2 OEM version comes complete with the PCB and all necessary parts, and costs about $30 for the kit form. (See Photo 1.) Assemble the module according to the directions from Parallax (www.parallax.com); however, do not install the bank of 16 angled male header pins on the lower edge of the board. This will permit direct wired connections between these BS2 pin connections and the keypad and control switches later.

There are several approaches you can use to assemble this project. You can breadboard the BS2 and related components if you intend on building the project as a temporary setup. Or, if you’d like a more permanent assembly you may want to house everything in a project case like I did.

I chose to sandwich the BS2 module and keypad on opposite sides of the project case lid. I soldered in a divided row of 16 female header pins on the output tabs of the BS2
module. For outputs marked Vdd, Vss, Vin, RES, P01, P02, and P03, install the female header pins on the top (component) side of the PCB. All remaining outputs (P04 through P15) had the female header pins installed on the underside of the PCB (See Photos 2, 3 and 4.)

At this point it’s a good idea to test out your BS2 module. First download and install the most recent version of the BASIC Stamp Editor (BSE) from www.parallax.com

- BASIC Stamp 2 OEM module kit ($30.95; Parallax.com, part # 27291).
- Velleman 12-key keypad (AllElectronics.com, part # KP-12; or from electronicpartsforless.com or www.willyselectronics.com, part # 12KEY); NOTE: The keypad must have separate connectors for each of the 12 keys, plus a common connector; a 3x4 matrix type keypad is NOT acceptable.
- (14) 10K ohm resistors, 1/4 watt.
- 220 ohm resistor, 1/4 watt.
- 1 μF electrolytic capacitor.
- (2) SPST toggle switches.
- Female breakaway headers, 16-pin strip (Sparkfun.com, SKU# PRT-00115); (Optional — only needed if using a project case and mounting keypad on outside.)
- SPST pushbutton switch, N.O. contacts.
- 9V battery clip with leads.
- External 8 ohm speaker with a connection cable, and 1/8 mono jack and plug for connecting with the encoder’s audio output.
- Solid hook up wire (and ribbon cable if using to interconnect keyboard and BS2 module).
- Breadboard (for temporary project mounting if not installing in a case).

You will need this in order to transfer programming code from your computer to the BS2 during construction, as well as to download your own modified code later on. Connect the positive terminal of the nine-volt battery to the Vin pin, and the negative terminal to the Vss pin. Connect the BS2 to your PC via either a DB9 serial cable or a supported serial-to-USB adapter. Once the BSE recognizes your module, you can test it out with a simple LED flasher circuit.

Connect the anode of an LED to pin 5 and connect the cathode to the Vss (ground) pin via a 220-ohm resistor. Then, download and install the LEDTest.bs2 program to the BS2 module (programs are available for download at www.nutsvolts.com). The LED should flash approximately once per second. If it doesn’t, check all connections on the module, as well as the LED test circuit, and fix any wiring errors. Once the module tests okay, remove the LED, 220-ohm resistor, and power connections, and disconnect the PC interface cable.

**Step 2: Add resistors.**

Once you have the BS2 module assembled and tested, solder a 10K ohm resistor to each of the pins labeled P1, P2, and P4 through P15 on the underside of the board. As shown in Photo 5, be sure to keep the lead from the board side of each resistor as short as possible. Solder the loose ends of these 14 resistors together via a short length of wire, forming a common connection to the BS2 module’s grounding pin (Vss). This is necessary in order for the microcontroller to function properly, as any switches connected to its I/O pins must be pulled low via resistors so that they are not floating when in the off (switch open) state.

**Step 3: Connect the keypad and switches.**

Now, connect the 12-key keypad to the appropriate BS2 module pin connections (see schematic.) This may be accomplished via ribbon cable or other point-to-point wiring. Use adequate lengths of wire between the
module and keypad, based upon how you will be mounting the finished project. Note that you must use a keypad with individual connections for each of the 12 keys and one common connection; I recommend using the Velleman 12-key keypad. Since I intended to house everything in a project case, instead of using ribbon cable I soldered a set of 14 male header pins to the underside of the keypad (i.e., the side without the keys). Photo 6 shows these header pins will connect to the BS2 module pins 4 through 15 on final assembly.

Next, solder an adequate length of hookup wire to pin 1 of the keypad to allow it to connect to Vdd on the BS2. The power switch is an SPST toggle connected between the nine-volt battery’s positive terminal and Vin. Connect the battery’s negative terminal to Vss. The DTMF/MF switch is an SPST toggle connected between P1 and Vdd. The Aux button is an SPST N.O. pushbutton switch connected between P2 and Vdd. The Audio output jack’s center conductor is connected to P0 via a 220-ohm resistor and the shield ring to ground. Connect the 1 μF electrolytic capacitor’s positive leg to the center connector of the Audio jack and the negative side to the Audio ground terminal.

Since I wanted to package the encoder inside a small (5” L x 3” W x 2” H) project case, I connected the keypad and BS2 module by sandwiching them on opposing sides of the case lid and joining them via their respective header pins through a slot I drilled into the lid. (I attached the back of the keypad to the outside of the lid with a plumber’s adhesive known as “Goo;” the BS2 module was supported inside the lid with a thin piece of Styrofoam and glued in place. Be sure that pin 3 of the keypad is connected to pin 4 of the BS2 module, continuing to connect the pins in a row so that pin 14 of the keypad connects with pin 15 of the BS2, leaving pins 1 and 2 of the keypad unconnected (see Photos 7 through 9). However you package the hardware, make sure the DB9 connector on the BS2 module will be accessible for programming. Assembly of the PhreakEncoder hardware is now complete (Photo 10)!

Now, finish the build by loading the software.

**Load the Code and Test It Out**

Connect the PhreakEncoder to your PC via the serial cable and use the BSE to load the PhreakEncoder.bs2 firmware (also on the NV website) onto the device. Connect a small eight-ohm speaker of your choice via the Audio output jack and turn on the power. With the DTMF/MF switch in the DTMF position, the keypad will produce standard telephone touchtones of 200 milliseconds duration. (If you do not hear tones in the speaker, check all wiring and connections against the diagram, and be sure
that the keypad and BS2 pins are correctly joined.) In the MF position, the PhreakEncoder will produce traditional MF/blue box tones formerly used on old telephone switching systems on keys 0-9 (which are needed by Asterisk's ProjectMF see Table 1).

When the Aux button is pressed, the keys produce alternate select tones per the chart in Table 2. Of course, these tones are for illustration purposes only. The magic of the PhreakEncoder is that you can easily change the open source BASIC Stamp program to produce any single- or multi-frequency tones you want! Sequences of touchtones can be programmed, turning the device into a quick dialer for phone, bank account, or other lengthy numbers you need to send via telephone. You now have an inexpensive, customizable tone machine which can generate unique tones for circuit testing or over-the-air radio control such as in amateur radio repeater control or other remote signaling purposes.

**Start Phreaking!**

The best way to use the PhreakEncoder is to hold the external speaker up to the telephone microphone, or near the microphone on a speakerphone (see Photo 11). Then, you would dial into one of the Asterisk Project MF numbers found on the Internet. Usually, the server will provide audio directions as to how to initialize its services.

**Hack the Code!**

The PhreakEncoder firmware is written using the PBASIC language. When examining the code using the BS2 IDE, pay close attention to those lines containing the DTMFOUT and FREQUENT commands. In order to alter the tones or produce tone sequences, you'll need to modify the arguments to these commands. The syntax for the DTMFOUT command is as follows: DTMFOUT pin_number, [digit1, optional_digit2, optional_digit3] — where pin_number is the desired output pin on the BS2, and digit(s) are the standard touchtone keys 0-9, *, and #.

The syntax for the FREQUENT command is as follows: FREQUENT pin_number, duration, tone1, [optional_tone2] — where pin_number is the desired output pin on the BS2; the duration is the tone duration in milliseconds; and tone1 and optional_tone2 are the desired audio frequencies in Hertz, with values between 0 and 32,767 Hz.

By reading the excellent Parallax support and reference materials available free on their website, you can pick apart the firmware and mod the source code as you desire. Take a look at Table 3 for a list correlating the BS2 output pins to keypad buttons. You'll need this in modding the code, to cross reference the keypad keys to the BS2 pins needed by the program. Don't forget that pin 3 on the microcontroller is unused in this project. (After reading the reference materials, who knows what mods you'll devise!)

**Why Build this Circuit?**

You may wonder why you should go through the trouble of building a dedicated circuit just to produce these tones for Project MF. If you remember the old school hacking/phreaking days, then the ability to recreate these activities within a legally safe environment is probably motivation enough. For those of you who weren't active in electronics at that time, you will be able to enjoy the same technical challenges and excitement experienced by hackers of a previous generation. And while not everyone who starts out building a blue box ends up founding a major computer company, you just never know. **NV**
Q & A ABOUT DIGITAL RADIO

I was talking to a friend of mine recently and he asked if I had ever built a crystal radio. I said yes, and asked why he wanted to know. He answered that he had never built one and just wanted to see if he could. I told him not to delay because pretty soon there would be no radio stations you could receive on a crystal radio.

Crystal radios are just a simple diode demodulator for amplitude modulation (AM) signals. Typically, crystal radios are built to receive local AM broadcast stations. And while those stations still exist, that might not be the case in the future. Why? Because almost all radios today have been converted to digital. AM broadcast stations are one of the few remaining radio services that still use analog AM. What ensued with my friend was a Q & A discussion of what digital radio is and how broad it is. Here is a summary of that discussion.

Q: What do you mean by the term radio?
A: A radio can be a receiver, a transmitter, or (more commonly today) a transceiver. Radio includes AM and FM broadcast, TV, any two-way radios, satellites, cell phones, hobby radios, marine and aircraft radios, military radios, radars, GPS, and so on. Anything we generally refer to as wireless means radio.

Q: How long has this digitization of radio been going on?
A: For decades actually. It probably started with satellites in the 1970’s, then moved on to other services over the years. Things really picked up in the 1980’s and I would guess that 90% of all radio is digital today.

Q: Just why did this move to digital occur?
A: Great question. And there is no one single reason. Probably the main reason is that digital signals are very robust and less affected by noise because of their binary nature. Second, we now have small, cheap, and fast integrated circuits and microcomputers that can process these signals. Designers found they could squeeze more signals into a given spectrum band with digital techniques thereby greatly improving spectral efficiency in a limited spectrum. Finally, some really great digital compression techniques were developed that further improve the ability to cram more signals at higher speeds into a limited bandwidth.

Q: Give me a good example of how some of those benefits were applied.
or (in some systems like GSM which is still in use), eight signals per spectrum channel. This really multiplied their subscriber capacity. Also created around this time was Code Division Multiple Access (CDMA) which further expanded the capacity of the cellular network. Today, all cell phones are digital and a wide range of different technologies are involved.

Q: Are two-way mobile radios also digital?
A: Some of them are but not all. Many like those used in fire, police, and public services still use FM. Digital technologies like P25 and Tetra have been around a while but the different services have been slow to convert. Military radios are mostly digital today.

Q: What about marine and aircraft radios?
A: Strangely, both are still analog. Marine radios use FM and aircraft radios use AM. I’m not sure of all the reasons for that, but I assume that since the ranges are short and the spectrum is not so crowded, why change? Besides, the technologies are well proven and the equipment is moderately priced and very reliable.

Q: What about broadcast radio?
A: Analog AM and FM stations are still around and still the most widely used. However, most stations have added a digital capability called HD radio. The station’s programs are digitized and broadcast in digital format on exactly the same frequency as the analog signals. The digital signals overlay the analog signals. Analog radios ignore the digital while an HD radio receiver can pick up the digital signals. The digital capability also allows most radio stations to broadcast two or three other programs simultaneously thanks to the multiplexing capability.

Q: Tell me more about digital HD radio.
A: Well, it has been around for years but in the past five years or so most stations have added it. I am not so sure that most folks actually know about it. You can readily buy an HD radio for about $100 at Best Buy, RadioShack, or any other consumer electronics retailer. Many cars come with HD or offer it as an option. The big deal about HD is that is gives slightly better fidelity than the AM and FM stations. AM stations sound more like FM and FM stations sound more like CDs. Furthermore, the HD signals are more immune to noise and do not fade as much when you are driving.

Q: Does the rest of the world have digital radio?
A: Yes, pretty much. Europe has had digital radio for years with a system called Eureka-147. You will hear it referred to as Digital Audio Broadcast (DAB). It broadcasts music and other programs in digital on the VHF bands (not the usual AM low frequency bands) allotted for that. Canada has a similar system.

Q: What about TV?
A: As you probably know, TV went all digital last year. All the analog stations went off the air and switched to digital which gave us high definition TV. If you still have an old analog set, you have to use one of those converter boxes that translates the digital TV signals to analog so your old set can handle them. All new sets sold today are digital only. Almost everyone loves digital TV as it has such improved resolution. Over 50% of the US population today has an HD set with a big LCD or plasma screen. There is also the mobile TV from FLO TV that you may have heard about. It was developed to provide digital TV to cell phones and portable/mobile TVs like those in cars and SUVs.

Q: Is digital TV unique to the US?
A: No, not really. Europe went to digital TV a number of years ago with its Digital Video Broadcast (DVB) standard. Japan has its ISDB (Integrated Services Digital Broadcast) system and South Korea uses their DMB (Digital Media Broadcast) system. Most of these digital TV standards have mobile versions for broadcasts to cell phones. The US is working on a mobile version of its Advanced Television Standards Committee (ATSC) standard for digital TV in the US. It should be available in the near future in some cell phones.

Q: Are there any radio services that have not moved to digital?
A: Yes, namely the hobby radio services. CB radio is still AM and SSB. The family radio service (FRS) is still FM. So is most ham radio which relies heavily on SSB and FM. Morse coded CW is a form of digital and that modulation is Amplitude Shift Keying (ASK) — a form of AM. Hams also use a form of digital called PSK31. These services lack the spectrum and the FCC restricts their modulation methods, but the movement is in the digital direction.

Q: What about shortwave radio?
A: Shortwave (SW) radio is still mostly analog. Virtually all worldwide stations broadcast in AM. It uses minimum bandwidth in a crowded spectrum but it is susceptible to noise as you know if you listen to SW.

Q: Will shortwave ever go digital?
A: Yes, and it is already happening. A digital standard called Digital Radio Mondiale (DRM) was developed and uses voice compression along with CODFM modulation. It is used primarily in Europe. The result is good quality audio that fits within the assigned narrow (20 kHz or less) SW channels. It's not wide spread yet, but it seems to be catching on. A number of radios are available in Europe but there are few on sale in the US. Some of the new digital software-defined radios have DRM demodulators.
Q: Are there any other digital radios you haven’t told me about?

A: Yes, there are almost too many to mention. For example, most short-range wireless data applications are digital. Some examples are garage door openers, remote keyless entry devices on vehicles, and remote temperature sensors. GPS receivers for navigation are also digital.

Q: What else?

A: Well, the Wi-Fi wireless LAN radios in your laptop or cell phone are digital, of course. These are the radios that let you access the Internet and your email at local hot spots. Bluetooth radios in your car or cell phone headset are all digital. Some wireless speakers for home stereo systems are all digital.

Q: You mentioned the HD radio in cars. What about satellite car radios?

A: Yes, satellite radios in cars are all digital. Sirius Radio and XM Radio merged recently under Sirius but they maintain separate digital satellite systems. Like other digital radios, the sound quality is CD level.

Q: What is an Internet radio? I’ve heard of that recently.

A: Oh yes, Internet radio. There is such a thing and it is not so well known. Basically, it is radio received over an Internet connection. Most radio stations also put their broadcasts on their website so you can go to that site and stream the audio to your PC or laptop. That means you need some kind of broadband Internet connection like a DSL line or cable TV access.

The interesting thing is that so much of what is on Internet radio is not generated by real radio stations. There are many non-radio stations that develop programs and content such as special music categories or talk sessions. I have heard estimates of 18,000 different so-called Internet radio stations.

Q: How do I receive these stations?

A: You can access individual stations on your PC if you have the speakers or headphones and know the URL. They also make special Internet radio receivers. A good example is the CC Wi-Fi radio shown in Figure 1 from C.Crane (www.ccrane.com). Crane makes high end AM/FM and SW receivers. This one connects by a Wi-Fi wireless link to your home Internet router and provides a way to access all those tens of thousands of stations. No antenna required.

Q: I had not thought of buying a separate Internet radio but it may be worth it to get that many stations. Is there any other way to receive Internet radio?

A: An unexpected method is to get Internet radio on your cell phone. Apple has several apps for streaming Internet radio to their iPhones. Weather Underground (www.wunderground.com) recently announced their WunderRadio — an app for streaming audio not only to the iPhone but also to BlackBerrys, Windows Mobile, and any phones using Google’s Android operating system.

Q: Wow, digital radio glut. Maybe I won’t build that crystal set after hearing about all my digital radio options. It may be more fun to explore one of these.

A: Go modern and start checking out all your digital radio options.
AVR Memory
Part 5: Bootloaders

Recap

We discussed bootloaders a bit in Smiley’s Workshop 22 (Busy as a BeAVR), and learned about the debt they owe to Baron Munchhausen’s bootstraps. This month, we’ll get more practical and build on last month’s AVR Memory Part 4 - Writing to AVR Flash where we learned how AVR Flash memory is structured and how to write to it. We tested that knowledge with the SmileyFlashWriter.c program. This month, we will build on this to create our own bootloader. You’ll need access to last month’s source code in Workshop26.zip [NutsVolts.com or SmileyMicros.com] to follow some of the instructions.

EduBootAVR - A Bootloader Written in C

A bootloader is a program that is used to download other programs. It usually resides in a special part of memory and is invoked when the device comes out of reset. It responds to data being sent – usually over a serial port – from an external device. The bootloader and the external device use a communication protocol that allows them to conduct transactions that mostly involve sending pages of binary program code to be written to the device application program section. For our demonstration, we will use avrdude on a PC to communicate with the AVR bootloader via the USART. As we will see shortly, avrdude is a very versatile tool that can handle many different programming protocols. We will be using the AVR109 protocol in avrdude but we will implement only those commands necessary for uploading program memory. There are other tasks that could be implemented such as allowing us to read and write EEPROM, but since we want to keep this bootloader small and simple, we won’t add those functions.

Deciding to Run the Bootloader or the Application

A bootloader is used to upload application programs. When you’ve uploaded an application, you then have two programs on your AVR: the application that begins at Flash memory location 0x0000 and your bootloader that begins at the NRWW (see last month) high memory location you’ve selected for the particular device (such as 0x3E00 for an ATmega328). Now, you have to decide which program you want to run when your AVR starts up. Generally, your system should be designed to run the application and only run the bootloader when you actually need it to upload some code. You could do this by having a function in your application that allows you to call the bootloader, but what if your application gets trashed? Then how will you upload your code? Atmel designed the AVR so that those with boot sections can have the device start up in that boot section if the BOOTRST fuse is set. With that fuse set, when the AVR starts after reset it loads the bootloader start address (also set by fuses) and starts the program from there.

The standard way to start up a microcontroller that is using a bootloader is to have the bootloader start and let it check for some indicator that it is needed. You could have the bootloader wait to see if an external programmer is trying to talk to it via the USART; after a brief pause with no attempt at contact, the bootloader can decide it isn’t needed and execute a jump to the application code at 0x0000. The downside of this technique is that you want the wait to be long enough to realistically allow a bootloader communication to begin, but not so long that a user might think something has gone wrong.
[Remember that folks using microcontrollers expect instant response and aren’t used to turning them on, then getting a cup of coffee while it boots like we expect with PCs.] Usually, we would check to see if the bootloader is needed in less than a second, and then switch to the application. Another method to tell the bootloader to stay active is to designate a pin state. You could, for instance, tell it to check PORTC pin 7 and if it’s high, run the boot code; if it’s low, run the application. This solves the time problem since it can be done in microseconds, but it adds the cost of designating a pin in hardware that has a switch or jumper to set its state. So, each method comes with a cost; one is time, the other is extra hardware.

Use Hardware to Select the Bootloader
You could have a hardware design that has a pushbutton connected to a pin that you intend to use for the application program. There is no reason that you can’t also use this to tell the bootloader to run. Just have the bootloader check that pin when it starts up out of reset. If it is pressed, then run the bootloader; if not, then run the application which can then take over the button for its own use since it is no longer needed by the bootloader. The following code snippet (which we do not use in EduBootAVR) would use a pushbutton that sets PORTD pin 7 to ground when pressed, or is pulled up to VCC when not pressed.

Place these definitions in the header:

```c
#define BOOT_STATE_PIN PIND
#define BOOT_STATE_PIN_NUMBER PIND7
```

Place this code in the bootloader main() function:

```c
// Use bootloader or application code? // If BOOT_STATE_PIN_NUMBER is low, // use bootloader.
if( !(BOOT_STATE_PIN & (1<<BOOT_STATE_PIN_NUMBER)) )
{
    // run the bootloader code
}
else
{
    // jump to the application code
}
```

Use Software Timeout to Select the Bootloader
We want to set some kind of timeout so that when the bootloader starts up on reset, it will wait for some short period. If nothing is talking to it in a valid protocol over the USART, then it will jump to the application section. This is what we use in our bootloader. We set a count based on the clock frequency divided by 16 (this is arbitrary and for the 16 MHz board would give a maximum count of one million). Then, we check to see if we have any new USART traffic. If not, we increment the counter until it reaches the maximum. A million may seem like a lot, but remember that this puppy is making this check about every clock cycle so it does it a million times in a fraction of a second. We run this count in the receiveByte function which calls application_ptr() to jump to the application if it counts out. Otherwise, it returns the received byte.

```c
uint8_t receiveByte( void )
{
    uint32_t count = 0;
    // Wait for data to be received
    while ( !(USART_CONTROL_STATUS_REG_A & (1<<USART_RECEIVE_COMPLETE)) )
    {
        if(count++ >= (P_CPU >> 4))
            application_ptr();
    }
    // Get and return received data from buffer
    return USART_DATA_REG;
}
```

The application_ptr() is not actually a function, but is a pointer to the first address in Flash, 0x0000, and is defined as:

```c
// Function pointer to jump to the application
// memory section
static void (*application_ptr)( void ) = 0x0000;
```

Okay, pointers like this are kind of weird, but don’t worry, we’ll discuss them eventually (if we haven’t already). For now, it works, so go with it.

Other Bootloaders
There are bunches of AVR bootloaders out there, and I can’t vouch for which is the best. Best is just too hard to judge. There was a long thread on www.avrfreaks.net [Google ‘smallest bootloader site:avrfreaks.net’]. On AVRFreaks.net, there are many bootloader projects around, but probably the two most discussed are ‘blips 4 bootloader’ by Steve Childress and ‘fast tiny & mega uart bootloader’ by Peter Danneger [Google these using site:avrfreaks.net]. There is a quite good avrfreaks thread [FAQ][C] Bootloader FAQ’ that discusses the pdf file: AVR Bootloader FAQ by Brad Schick and Cliff Lawson. I’m not providing links because avrfreaks links are very long, so you’ve got some Googling to do.

Please note that the bootloader we are designing for this article won’t work with the Arduino using the Arduino IDE. It works just fine for the Arduino board with avrdude and the script provided here, but the Arduino IDE is hardwired to use an ‘avr isp’ type bootloader. If you use this bootloader with an Arduino then want to restore the original bootloader, you can get a copy at:

www.ladyada.net/library/arduino/bootloader.html.

The AVR109 Bootloader Protocol
Atmel has promulgated several bootloader protocols – as have others – and again I can’t vouch for which is the best, but I’ve chosen AVR109 [www.atmel.com/atmel/acrobat/doc1644.pdf]. This application note is accompanied by the source code written for the IAR compiler. I haven’t bothered to port it to gcc since I’m only interested in a subset of the commands available. This protocol is the AVR side of the equation and the PC side is covered by the application note: AVR911 Open Source
Programmer. It is compatible with the AVRProg utility that comes with AVRSTUDIO [I use AVRProg for the Butterfly in my C Programming book, but for this Workshop we'll use avrdude which also speaks AVR109]. This protocol covers reading and writing Flash and lockbits and EEPROM and reading Fuse Bits, but we will only look at the Flash.

**AVR109 Commands We Use**

The heart of our bootloader is the function `AVR109CommandParser()` shown next. This function contains a switch statement that calls one of the AVR109 commands, depending on the command byte received over the serial port. Not all the AVR109 commands are used since only those chosen are needed for a simple bootloader. Most of the function names are the same as the command name and are somewhat self explanatory [autoIncrementAddress()] is called by the command byte 'a' which tells the bootloader to auto increment the address — whoa, this code is almost self-documenting!]. [This is a joke since anyone who claims their code is self-documenting is wrong and should be ignored from that point forward (as should their code). They are also dangerous (as is their code), but we'll save that for another rant.]

```c
// Note that the function name is the command // except where otherwise commented.
void AVR109CommandParser()
{
    uint8_t cmd,
    while(1)// Loop forever
    {
        cmd = receiveByte();
        switch (cmd)
        {
        case 'F': // Enter program mode
        case 'L': // Leave program mode
            sendByte('\n');
            break;
        case 'a':
            autoIncrementAddress();
            break;
        case 'G':
            returnSupportedDeviceCodes();
            break;
        case 'S':
            returnSoftwareIdentifier();
            break;
        case 'G':
            returnProgrammerType();
            break;
        case 'b':
            checkBlockSupport();
            break;
        case 'B': // Start Block Flash Load
        case 'G': // Start Block Flash Read
        startBlockAction(cmd);
            break;
        case 'E':
            exitBootloader();
            break;
        default:
            if(cmd != 0x18) sendByte('?');
            break;
        }
    }
}
```

Some of these functions are required by avrdude when running in the AVR109 mode, though several don't do much and wouldn't really be needed if you want to write your own PC side bootloader and use it with a device that you already know about. You wouldn't need to get the identifying info such as the signature byte, the programmer type, nor would you need to see if it supports blocks. Neither would you need to tell it to enter or leave the programming mode since what else would it do?

In that latter case, all we do is return a "y" that lies to avrdude so that it will continue working. Why write our own when avrdude is a great and free program? I'm not going to try to write another version of it just to save a few bytes of code. Remember that this bootloader just barely gets in under the 1,024 byte limit and the next lower limit is 512 bytes, so we'd have to eliminate half the code to have a chance of getting there (and that isn't likely). If you really need to go that low, check out some of the assembly language bootloaders and PC side programs that communicate with them.

The critical action takes place in the `startBlockAction(cmd)` function:

```c
void startBlockAction(uint8_t cmd)
{
    uint8_t tempRec;
    uint16_t tempSize;
    tempSize = (receiveByte() << 8)
             | receiveByte();
    // dummy read for type - we only do flash
    // NOT EEPROM
    tempRec = receiveByte();
    if(cmd == 'B') blockFlashLoad( tempSize );
```
else blockFlashRead( tempSize );
}

This function – depending on whether cmd is 'B' or 'G' – calls functions to load or read the Flash. The load function is pretty much the same as the Flash load function we discussed last month for the SmileyFlashWriter. The blockFlashRead() function uses the avr libc pgm_read_byte_near(address) function similar to what we saw in Workshop 25.

void blockFlashRead(uint16_t size) {
    uint8_t data;
    do {
        // read_program_memory(address, 0x00);
        data = pgm_read_byte_near(address++);
        // send byte
        sendByte(data);
        // reduce number of bytes to read
        // by one
        size--;
    } while (size);  // loop through size
}

Keeping the Bootloader as Small as Possible

When we compile EduBootAVR with AVRStudio/WinAVR with our usual settings, we get a code size of 1,086 bytes. Oh, darn, the boot section boundary is at 1,024, so it is 62 bytes too large. Can we get the code smaller? Could we rewrite the program hoping to find something that will save us 62 piddly bytes? Well, we might, but there are other ways to save some space with a bootloader.

Eliminate Some of the Start Code

When we use AVRStudio with WinAVR to compile files, we get a gift of having necessary startup code added for us. This includes the interrupt jump table. But what if you aren’t using the interrupts? Well, you can eliminate it. Since we aren’t using interrupts in the bootloader, let’s dump the jump table! We do this by opening the Project Configuration Options window Custom Options pane, as shown in Figure 2. We write ‘-nostartfiles’ in the ‘Add’ test box, then click ‘Add’ so that ‘nostartfiles’ appears as shown in the figure. This also removes some startup code that we do need, so we have to add it back by putting the following code snippet right before the main() function:

// From Peter Flury AVR Freaks Aug 10 2005 -
// to remove interrupt Vector table
// put -nostartfiles in LDFlags, add the
// following function saves wasted space
void _jumpMain ( void ) __attribute__( (naked) ) __attribute__( (section (".init9")));
void _jumpMain ( void ) {
    asm volatile ( "set __stack, %0 :: "i" [RAMEND] );
    asm volatile ( "clr __zero_reg" );
    // rl set to 0
    asm volatile ( "ljmp main" );
    // jump to main()
}

Now when we compile the code, we get 958 bytes which will fit in the desired boot section. We could stop here, but there is also another way to save some space.

Use a Different Optimization Level

We can keep the code unchanged and instead of using the default optimization of -Os, we can use -O1. You change this as shown in Figure 3. Just select -O1 from the Optimizations drop box. When you compile using this optimization, you get 986 bytes which would also get you inside the limits.

If you use both, you get 854 bytes, providing you with
even further headroom. There may be even more ways to save space, but we are already down 170 bytes, so why bother? If you really want to get it smaller, check out the assembly language bootloaders mentioned earlier.

EduBootAVR Source Code

The source code for this month's article is located in Workshop27.zip as usual, and — not so usual — it is also located on Google Code Project Hosting: https://code.google.com/p/edubootavr/source/browse/#svn/trunk.

This is my first attempt to provide a Collaborative Open Project — a concept that we'll dive into next month.

Using the STK-500 to Program the ATmega644

We will use the STK-500 to program our ATmega644 chip. The setup is shown in Figure 1. I strongly suggest that you read AVRStudio Help\AVR Tools User Guide; click on the STK-500 and read about it. You'll see that there are many things that the STK-500 can do, but we'll just use it to do ISP programming on an ATmega644 chip in a socket on the board. In Workshop 22, we saw how to use the Dragon to program an ATmega644 on a breadboard (BeAVR40 design) using a hardware setup that would make Medusa look cute. If you got that working, then you can just follow the directions there and forget about the STK-500.

However, if you have access to an STK-500, you may have an easier time since you can plug the ATmega644 into a socket and then use a 12-pin cable that comes with the STK-500 to program the chip (thus saving Medusa for the Kraken). It is much simpler to set up and therefore less problem-prone than the other way.

In AVRStudio, click the Connect Dialog as shown in Figure 4. In the Select AVR Programmer window shown in Figure 5, highlight STK-500 and Auto, and click Connect. (If it complains that it can't find the STK-500, make sure that you are using a COM port in the acceptable range.)
In the Programmer Main tab, read the signature just to verify that you really are communicating with an ATmega644 (as shown in Figure 6).

In the Fuses tab (see Figure 7), set the BOOTSZ to $7E00 and check the BOOTrST fuse.

Finally (as shown in Figure 8), open the Program tab and make sure you are pointed to the hex file; then click Program.

Using EduBootAVR

Great! Now we have our own bootloader that we can use to upload programs. Let’s take the ATmega644 that we just used the STK-500 to load the bootloader and put it on a breadboard like the BeAVR shown in Figure 9 (discussed in detail in Smiley’s Workshop 22 – Busy as a BeAVR). In earlier Workshops, we’ve seen how to use bootloaders with various programs. This includes AVRProg with the Butterfly; the Arduino IDE with the Arduino; and avrdude with several things. We’ll finish off this month by using avrdude in the Command Prompt. We’ll do this cookbook style since we’ve seen the details before.

- Make sure that you have the code from Workshop 25 located at: C:\pgmtst. We will upload this to our application section using edubootavr.
- Make sure that you have compiled pgmtst for the ATmega644 (BeAVR40).
- Find the Command Prompt program (probably under Start\All Programs\Accessories\Command Prompt).
- Open Command Prompt and type in: cd C:\pgmtst\default (if that is where you’ve put that code) and hit return to take you to the pgmtst default directory where pgmtst.hex resides.
- Find the COMx port your BeAVR40 is talking to.
- Open Notepad and type in:
  avrdude -p m644p -c avr109 -P COM3 -b 57600 -v -F -D -U flash:w:pgmtstBeAVR.hex.
- Copy that and paste it to the Command Prompt as shown in Figure 10.
- Hit enter and it should upload pgmtst as shown in Figure 11.
- Open Bray’s Terminal and reset the BeAVR40 which should show the text in Figure 12.

So, now you know all about AVR memory. If you thought this was edifying, then don’t forget to check the Nuts & Volts WEBSTORE section in the magazine or on the Internet where you can get my books and project kits to help you learn and have some fun doing it. NV

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**Figure 12. Bray’s Terminal.**
IMPLEMENTING AN ADC KEYPAD

Way back in March '07, well before the PICAXE Primer appeared on the scene, I wrote an article for *Nuts & Volts* that described how to interface a 12- or 16-key matrix keypad to a PICAXE-18X processor. In the article, I used a standard matrix-scanning approach to decoding a keypress. For a 4 x 4 matrix keypad, this approach requires eight I/O lines (four inputs and four outputs) for the interface. That's a large number of I/O lines to dedicate to a single function, but I didn't mind the I/O "expense" because the keypad was being used as part of a stand-alone I/O terminal and the 18X processor had more than sufficient resources for the task. However, I'm sure you can think of several 08M-based projects that would also benefit from user input via a keypad, so this month we're going to focus on a different approach to interfacing a matrix keypad with a PICAXE processor. This alternate approach only requires a single ADC input to decode a 12- or 16-key matrix keypad, so it's suitable for use with any 08M-based project. Of course, the same approach can be used with any PICAXE processor, as well.

OLD BUSINESS

Before we get into the details of this month's project, I have a small piece of "old business" to discuss. Back in the April '10 installment of the Primer (the one in which we developed software for our MAX7219 LED Display Driver), I mentioned a problem I had in getting the driver software to function correctly. (See the "Testing the Serialized LED Display" section that starts on page 18 of that article.) I initially tried to "blank" the four LED digits by using the following code snippet:

```c
for maxreg = 1 to 4
  outbyte = blank
  gosub shout
next maxreg
```

The problem was that this approach didn't work, but I couldn't figure out why, and my publication deadline was rapidly approaching. To "solve" the problem, I took the easy way out and modified the code so that it did work, but I did invite readers to figure out what was going on.

Fortunately, I received an email from an astute reader (Matt Rusnak) that clarified the cause of my little problem — thanks Matt! In effect, I had ignored my own advice.

In the previous Primer column (February '10), I reminded readers that whenever a word variable is declared (e.g., `symbol outword = w0`), it's important to not also declare either of the associated byte variables (e.g., `b0` or `b1`) because changing the value of one variable (e.g., `b1`) will inadvertently also change the value of another variable (w0, in this case).

I then went on to explain why our MAX7219 project was an exception to this rule (see the original article for the explanation), and to make the following declarations in the driver software:

```c
symbol outbyte = b0
' data to be transmitted
' to the MAX7219
symbol maxreg = b1
```

So far, so good, but the problem entered the picture when I forgot the fact that the `shout` subroutine modifies `outword` (w0). Therefore, it also modifies `maxreg` (as well as `outbyte`, but that's not a problem), which causes the above `for` loop to misbehave.

As usual, once the cause of the problem is clearly understood, the solution is relatively simple. All we need to do is declare an additional byte variable to use in the `for` loop. (Of course, we can't use `index`, because the `shout` subroutine uses that variable as well.) It's a little embarrassing to have been "hoist by my own petard," as they say, but I do appreciate the feedback and thought it worthwhile to pass the information on to everyone. It certainly emphasizes the need for caution.
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when declaring “overlapping” word and byte variables.

NEW BUSINESS

There’s another bit of information I want to share with everyone. I'm pleased to announce the recent publication of my first book, PICAXE® Microcontroller Projects for the Evil Genius. By the time you read this, it will be available at the NV Webstore (http://store.nutsvolts.com). In addition to simply being a shameless plug for the book, I'm also mentioning it here because I need to include the following statement: Portions of the following project are excerpted from the book with permission from McGraw-Hill.

INTERFACING A 4 X 4 MATRIX KEYPAD

Actually, the Evil Genius book includes three different projects that involve a matrix keypad. We're going to be focusing on a variation of the first project, but when we have finished I'll also include some information on what we did next in the book. We'll be using the same keypad from the book's projects (a 4 x 4 matrix), but you certainly can adapt the project for use with a 4 x 3 matrix keypad if you prefer.

The standard arrangement for a 4 x 4 matrix keypad is shown in Figure 1. Whenever a key is pressed, the connections for the corresponding row and column are shorted together; for example, pressing the “6” key connects the pins for row 2 and column 3. You don't need to use the same keypad that I did. Any matrix keypad you have on hand should work just as well — just be sure that it is in fact a matrix layout. I have seen keypads that look similar to a “genuine” matrix, but actually have each key connected to its own line with a common connection to all the keys. It would be possible to use this type of keypad, but the necessary modifications would be extensive.

The keypad that I am using includes a row of eight holes along its bottom edge, into which a male header can be soldered so that the keypad can easily be connected to a breadboard, stripboard, or ribbon cable assembly. For breadboard projects, it would actually be more convenient to have the header at the top so that the keypad could be placed in front of the breadboard rather than behind it. If you prefer that arrangement, you may be interested in the second Evil Genius keypad project in which we perform a little “surgery” on the keypad (more about that later). For now, we’ll stick with the stock configuration.

There seems to be a fair amount of variation in the specific order of the connecting points on different matrix keypads. The simplest arrangement for the user would naturally be to have the row and column connections in order. However, probably because the keypad layout is simpler and therefore less expensive, the connections are frequently not in any logical order. Figure 2 shows the pin assignments for the keypad that I used (which is available on my website at www.jrhackett.net). If you use a keypad that has a different pin arrangement, you will need to modify the breadboard layout. The schematic, of course, will remain the same.

USING AN ADC APPROACH TO DECODING MATRIX KEYPADS

The ADC approach to decoding a matrix keypad employs a resistor matrix to produce an analog voltage level that differs for each possible key-press. (The output voltage remains at ground level as long as no key is pressed.) The analog voltage is connected to an ADC input and the processor simply converts the resulting voltage level back to the corresponding keypress. Figure 3 presents one possible arrangement for a resistor matrix that can be used in the ADC approach to keypad decoding.

As you can see, each possible key-press connects one, two, or three resistors in series between the +5V line and ground. By carefully choosing the values of the seven resistors in the layout, it's possible to produce 16 voltage levels that are spread out far enough to easily differentiate. I would like to be able to say that I used a sophisticated mathematical formula to determine the required resistor values, but I actually used a more primitive “trial
and error” approach to the problem.

Before I discuss my method and the results I obtained, I need to mention an important factor to keep in mind. Standard 1/4 watt resistors have a 5% tolerance rating which means that a 10k resistor can actually measure anywhere between 9.5k and 10.5k. This is why it’s important to be able to produce a wide range of analog voltages. If the ADC readings for two adjacent keys were too close to each other, variations in actual resistor values could result in misidentifying the specific key that has been pressed.

In order to make my trial and error approach as painless as possible, I set up a simple Excel spreadsheet to compute the ADC values that would result from a specific combination of resistors and then tried various combinations until I found one that worked. When I used the 256 levels provided by the readadc command, some of the ADC values for two adjacent keys were so close (differences of 4 or 5) that I was concerned that it might result. Switching to the 1024 levels provided by the readadc10 command greatly simplified the task.

The resistor values that I finally chose are shown in Figure 4. Each of the 16 “key” positions includes two pieces of relevant data: the total resistance that is connected in series with the 10k base resistor when the corresponding key is pressed; and (in parentheses) the resulting value produced by the readadc10 command.

If you would like to experiment with different resistor values, the formula you need is the basic voltage divider rule. In English: The ADC reading is to the maximum ADC value (1023 for readadc10) as the base resistance (10k) is to the total resistance. Figure 5 presents the same thing algebraically if you prefer it that way.

If you double-check some of my computations,

**CONSTRUCTING AND TESTING THE BREADBOARD CIRCUIT**

The schematic for our breadboard circuit is shown in Figure 6. As I mentioned earlier, the eight connections to the keyboard that I’m using are not logically ordered, but it really doesn’t matter much. The important thing is to make sure that the connections for each resistor are the same as the ones presented earlier in Figure 2. If the pinout is different for the keypad you intend to use, simply rearrange the connections appropriately. The parts list for our breadboard circuit is too simple to warrant a separate sidebar; just a matrix keypad with a male header, the seven resistors shown in the schematic, and a PICAXE-08M processor are needed.

My breadboard layout is shown in Figure 7. To save some space on the breadboard, I’m using the Axmate-FT programming adapter which is also supplying power to the breadboard. The small red printed circuit board (PCB) attached to the Axmate-FT is the 5V version of SparkFun’s latest FTDI Basic Breakout board (DEV-09716) which I have reconfigured to work correctly with PICAXE processors (as we discussed back in the June Primer). I really like SparkFun’s new board because they have moved the six-pin female header underneath which makes this the smallest FTDI-based board that I have found so far.

When I assembled the Axmate-FT in the photo, I installed its six-pin male header on the bottom of the board so that the two boards line up correctly. (If you’re not using an Axmate-FT adapter, you will need to install all the parts farther to the right on the breadboard to allow enough room for a power supply and programming adapter.)

When you have assembled your breadboard circuit, we’re ready to discuss our first program (keypad1.bas). Download it from the Nuts & Volts website (www.nutsvolts.com), along with keypad2.bas which we’ll discuss
shortly. (At this point, you may want to print out both programs for reference throughout the following discussion.) Keypad1.bas has two purposes: to make sure that the breadboard circuit is configured correctly, and to determine how closely our “real-world” ADC values match the theoretical values that were presented back in Figure 4. The main program loop repetitively carries out the following tasks:

- Wait for keypress. (Theoretically junk = 0 if no key is pressed, but “< 5” is safer.)
- Pause for 50 ms to de-bounce the switch. (The pause 100 is halved at 8 MHz.)
- Get ADC value (using readadc10).
- Wait until switch is released.
- Send the digits of the value to the terminal window (followed by CR and LF).

When you have downloaded keypad1.bas to your OBM, pressing any key should produce a value between 0 and 1023 in the terminal window. If it doesn’t, you’ll need to troubleshoot your breadboard setup. When your circuit is functioning properly, jot down the value you obtain for each keypress. We’re going to use these results soon, when we get to the keypad2.bas program. Be sure to press each key a few times because you may get slightly variable readings for some of the keys. If so, just choose the most typical value for each key and make a note of it. (As I mentioned earlier, your results will most likely differ somewhat from the theoretical values back in Figure 4, but they should be close.)

### DECODING THE KEYPRESSES

Now that you know the ADC value that’s produced by each keypress in your breadboard setup, we need to modify keypad1.bas so that it actually decodes each keypress and outputs the appropriate character. Naturally, we’ll need another variable in which to store the resulting character; let’s call it char. Also, we’re going to use a select case statement to accomplish the decoding, but we can’t use a series of equalities to convert each ADC value to the correct character because (as we just discovered) some of the values are slightly variable. Also, additional slight variation can be introduced by changes in temperature and other factors. To make sure that we always decode the correct character, we’ll use a series of “less than” phrases in the select case statement and work our way up the list. Using this approach, our select case statement will take the form of the following code fragment:

```code
select case key
  case < 369 : char = 49  ' ASCII code for "1"
  case < 385 : char = 50  ' ASCII code for "2"
  case < 402 : char = 51  ' ASCII code for "3"
  case < 434 : char = 55  ' ASCII code for "A"
  etc., etc.
end select
```

The question remains: What specific values should we use in place of the question marks? The safest (i.e., most error-free) choice is the mid-point between the values for each pair of adjacent keys. I’ll use our theoretical values to clarify this point, but you should substitute the actual values you obtained from running the keypad1.bas program. The theoretical value associated with the “1” key is 361 and the value associated with “2” is 376, so the mid-point is about 369. Therefore, in our select case statement, we’ll say that any value less than 369 will be decoded as the “1” character. If we use the same approach to each of the characters, our select case statement becomes:

```code
select case key
  case < 369 : char = 49  ' ASCII code for "1"
  case < 385 : char = 50  ' ASCII code for "2"
  case < 402 : char = 51  ' ASCII code for "3"
  case < 434 : char = 55  ' ASCII code for "A"
  etc., etc.
end select
```

Since our select case statement involves 16 distinct cases, it will be fairly long. To compress it a bit, we’re going to use a little shortcut. Similarly to many dialects of BASIC, PICAXE BASIC supports the use of the “:” symbol to separate multiple statements on the same line which we can write:

```code
select case key
  case < 369 : char = 49  ' ASCII code for "1"
    case < 385 : char = 50  ' ASCII code for "2"
    case < 402 : char = 51  ' ASCII code for "3"
    case < 434 : char = 55  ' ASCII code for "A"
    etc., etc.
end select
```

The “:” shortcut is not something I’m suggesting you use frequently in your programs; it can easily make code much more difficult to read. However, our long select case statement is perfectly readable in this form, and also much shorter. One final point – don’t forget, as soon as one of the case conditions evaluates to true, the associated code is executed and the remainder of the select case statement is skipped. For example, if key equals 375, char is set equal to 50 and the compiler jumps ahead to the program line that follows the endselect statement.
Our second program (keypad2.bas) incorporates all the changes we just discussed. Open it in the Programming Editor and change the theoretical mid-point values in each of the case conditions to the values you calculated from the results of running keypad1.bas. Also note that the servd statement has changed; instead of sending the individual digits of the ADC value, we’re now transmitting the appropriate ASCII value for each character to the terminal window. Download the program to your breadboard circuit and test all the keypresses. You should see the correct character appear in the terminal window in response to each keypress. If not, a little troubleshooting is in order.

WHAT’S AN EVIL GENIUS TO DO?

That’s as far as we’re going to take our ADC keypad project this month. I’m sure it’s no surprise that the project we tackled in the Evil Genius book was to design a serial version that simplified the interface with a master processor by freeing the master from the chore of repetitively monitoring an ADC input. Next, we performed a little “surgery” on our keypad. We disassembled it and rearranged its innards, so that the connector was moved to the top of the keypad and its keys were shuffled into a more convenient pattern. The results of our little operation are shown in Figure 8.

As you can see, it looks like four of the keys have been replaced, but actually the arrows were printed on a Brother PT-300 label maker and glued onto the four keys. I used black tape with white letters so that I could completely cover the original characters on the keys. Also, I deliberately retained the “B” key as a “Back” key in a menu structure, and the “A” key as an “Accept” (“Enter”) key. I realize the new key layout results in a slightly unusual arrangement for the 10 digit keys, but I think this compromise is justified by the convenient layout for the four arrow keys and the fact that the “Enter” key is in the lower right-hand corner where it should be.

Next, the modified keypad was used to develop a programmable, multi-function peripheral device (MPD) that can be easily connected to any breadboard circuit (see Figure 9). Because the MPD is fully programmable, it can be used to implement a variety of helpful utility and troubleshooting applications for use with any PICAXE project. The following are just a few of the many possibilities:

- Stand-alone I/O terminal with sound
- Digital frequency generator
- Frequency measurement of project signals
- Timing of project routines and waveforms

Because the MPD is based on a PICAXE-20X2 circuit, it can implement any of these functions with a surprising amount of speed and accuracy. In addition, the 20X2 has enough memory to store several programs at once. In fact, we also developed a simple operating system (MPDOS) that makes it easy to switch from program to program as the need arises.

COMING ATTRACTIONS

In addition to the publication of PICAXE® Microcontroller Projects for the Evil Genius, there’s another major event that’s about to occur in the PICAXE world. Revolution Education has announced its newest processor — the PICAXE-18M2 — and it’s scheduled to be available by the time you’re reading this column. The 18M2 is a revolutionary product with many impressive hardware and software enhancements. I don’t have space this month to get into the details, but take a look at Figure 10. I think you’ll be impressed.

The 18M2 also includes several software enhancements; I’ll just mention two of them to whet your appetite. There’s a new built-in time variable that keeps track of elapsed time in the background while your program is carrying out other tasks. This means that a project will be able to include a real-time clock feature without necessitating additional hardware or complicated software interrupt routines. Also, the 18M2 will be able to run up to four different program tasks in parallel at the same time by rapidly switching among the various program “threads.” This new capability isn’t multitasking in the strict sense, because program instructions are still being executed one at a time. However, the automatic switching among program threads will occur so rapidly that it will create the illusion of multitasking.

I have saved the most impressive specification of the new 18M2 processor until last; it’s slated to retail for about the same price as the older 18M processor. To be able to purchase that amount of processing power for less than $4 is truly amazing. I’m eager to get my hands on one, and as soon as I do (which will almost certainly be in time for the next Primer column), I’ll get to work on reporting what I learn about this impressive new chip. See you then...

| FIGURE 10. Comparison of 18M2 vs. 18M selected features. |

<table>
<thead>
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<th>Feature</th>
<th>18M2</th>
<th>18M</th>
</tr>
</thead>
<tbody>
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<td>256</td>
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<td>General Purpose Vars</td>
<td>28</td>
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<td>Max Operating Speed</td>
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</tr>
</tbody>
</table>

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- **Subscriber's Price:** $37.95
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**Talking Skull Kit**

Get ready for a tech-scary Halloween! New for 2010, the audio board is now assembled!

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Available in blue, black, red, and green. All components are pre-cut and pre-bent for easier assembly and the microcontrollers are pre-programmed with the software. Kits also include PCB, AC adapter, and instructions on CD-ROM.

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- **Reg Price:** $195.95
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Hydrocar Kit

The Hydrocar is used in a couple of great projects from the series of articles by John Gavlak, "Experimenting with Alternative Energy." In Parts 10 and 11, he teaches you the operation of the Polymer Electrolyte Membrane "reversible" fuel cell. For kit details and a demo video, please visit our webstore.

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The Windpitch Wind Turbine Kit is a miniature real-working wind turbine and is one of the great projects from the series of articles by John Gavlak, "Experimenting with Alternative Energy." In Parts 8 and 9, he teaches you how to produce the most power by evaluating the pitch (setting angle) of the profiled blades. For kit details, please visit our webstore.

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The function of a Silicon Controlled Rectifier (SCR) is to conduct electric current only after a positive pulse is put on its gate pin. The SCR will continue conducting until power is removed from the cathode and/or anode. SCRs are typically used in alarm system circuits.

1. **Build the Circuit.**
   Using the schematic along with the pictorial diagram, place the components on a solderless breadboard as shown. Verify that your wiring is correct.

2. **Do the Experiment.**
   **Theory:** In this simple circuit, the electrons will flow from the battery through the SCR, through the LED, and back to the battery only if the SCR receives a positive pulse on its gate. The positive pulse is put there through the 1,000 ohm resistor connected to the positive terminal on the nine-volt battery. (A 1,000 ohm resistor is used to make sure that not too much current flows to the gate.)

**Procedure:**
Connect a nine-volt battery to the battery snap. Touch the wire coming from the 1,000 ohm resistor to the gate of the SCR. This positive pulse from the battery will cause the SCR to conduct and light the LED. It will remain lit even after you remove the pulse from the gate.
RESISTORS

R1 = hole 10a  hole 23a
220 ohms, (Red, Red, Brown, Gold)
R2 = hole 14d  hole 23d
1000 ohms, (Brown, Black, Red, Gold)

LIGHT EMITTING DIODE (LED)

anode  hole 10c
led  hole 9c

BATTERY red = hole 23a
black = hole 20f

WIRES W1 = 6j and 20j
W2 = 9d and 5i
W3 = 14e and (touch to gate of SCR)

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<<< QUESTIONS

Reactive Transformerless Power Supply

How does one calculate the value for the reactance capacitor in an AC to DC reactive power supply so as not to over-work the zener diode? I have used from 0.68 µf to 4.7 µf for various voltages out, but it has always been trial and error.

#10101 Michael Stevens
Rogersville, AL

Ping Tester

I work in large building complexes and need to "ping" devices several floors away or in the next building. I typically use a laptop, but it's difficult to hold or set down while in riser or mechanical rooms. There are cable testers with ping functionality but they are costly. It would be a neat project to build a ping tester using one of the widely available development boards and microcontrollers. There are so many choices I don't know where to start. Could someone point me in the right direction?

#10102 Jim, via email

Adapter Board

I recently got a Needham PB-10 EPROM programmer. It can program 4751H microcontrollers, but requires an adapter board to do so. I can't find any info on this board. Does anyone have a schematic or info so I could roll my own?

#10103 Pierre Olivier, Quebec

Resistance Inverter/Converter

I installed a dash from a Buick into a Winnebago. The gas gauge in the new dash requires 242 ohms for a full and 42 for an empty reading. The Winnebago sending unit gives a 10 ohms full and 180 ohms empty resistance reading. I need help in designing a circuit to do this conversion.

#10104 Kenneth Lysek
Henderson, CO

Racing Game Data Logging

I want to connect an analog tachometer and speedo or an instrument cluster to a PC via USB for receiving data from the Live For Speed racing game.

What Arduino and circuit components do I need to use on this project and which software is more useful?

#10105 Ozgur
Istanbul

<<< ANSWERS

[#4105 - April 2010]

Steampunk Keyboard

I am building a Steampunk keyboard for my own use. I would like to have a small motor spin some gears each time a key is pressed on the computer keyboard. I have not been able to locate an encoder output or an activity monitor that will show key presses. I don't care which keys are being pressed, and the gears spinning are just for show. I would like a hardware solution over a software solution — but a complete *.exe program would be acceptable. (Maybe flashing the scroll lock LED when keyboard activity is being transmitted to the keyboard controller.)

Monitoring the data stream for F0 (Key up) occurrences would also work.

The circuit diagram in Figure 1 should do the trick; just connect 'IN' to the keyboard's data line. The PS/2 data line is normally held high, but is guaranteed to be pulled low at the beginning of a transfer. A simple monostable 555 can detect that low pulse and lengthen it to provide a control for your motor.

With the indicated values, the 555 should run the motor for 0.1s - 1.2s (depending on the setting of R2). I have included two options for connecting your motor. Option 1 should work for small motors that draw less than 0.2A and operate at 5V.

If you have any other configuration, I recommend Option 2. If you feel that the time isn't long/short enough, just change the values of R1, R2, and C1 as needed (Max time = \((R1 + R2) \times C1 \times 1.1\); Min time = \(R1 \times C1 \times 1.1\)).

Daniel De Jager
Edgewood, WA

[#6101 - June 2010]

Floppy Interface

I am looking to buy or get enough tips to design and build a device to do service adjustments on old style floppy drives that are used in music keyboards. To run analysis and adjustment software, I have to remove the drive from the keyboard and connect it to a computer. But, the new computers don't have floppy connectors any more.

By readers and no guarantees whatsoever are made by the publisher. The implementation of any answer printed in this column may require varying degrees of technical experience and should only be attempted by qualified individuals.

Always use common sense and good judgment!
more. The best device would have a floppy drive 34-pin connector on one end, a USB plug on the other, and some controller in the middle. The current USB external floppy drives are quite different from anything I could use for this project.

#1 The best way to approach this and probably the least problem-filled is to get an older PC — one that will run your adjustment software (lots of PIDs running XP out there that you can buy for a song). Get a long 34-pin floppy cable, as well as a long power cable extension, open the PC up and connect the floppy cable to the motherboard, connect the power cable extension, pop one of the front covers off of the PC, snake the cables through the opening, then seal the PC back up. Most PC motherboards still make a provision for a floppy drive connection so that isn’t a problem. A search on the Internet for the cables (try CyberGuys - www.cyberguys.com or Electronix - www.electronix.com) shouldn’t be difficult. Now you have a test PC with the necessary cables hanging out the front of the unit. Unfortunately, they do not make an external floppy adapter like they do for making external IDE or SATA devices that use USB connectivity. What makes this setup nice is if you can connect to the Internet with the PC, not only do you have a nice test jig you also have a PC to surf the Internet for your repair business.

Ralph J. Kurtz
Old Forge, PA

#2 You did not specify 3.5, 5.25, or 8" floppy drive. Here is something that should work:

The Buslink FDD1 3.5" USB floppy drive is a standard 3.5" floppy drive with an easily accessible 34-pin ribbon cable in a case with a USB to floppy controller. Will it work with a 5.25" drive? I don’t know yet, but it’s the first time we have found a 3.5" USB floppy that isn’t "integrated" and that consists of a USB to floppy controller with a standard 34-pin interface. Several places sell these in the $40-$60 range. Try www.pcruish.com/product/Floppy-Disk-Drives/590364/Buslink-FDD1-Floppy-Drive or at Amazon Market place from www1.shopping.com/xPO-Universal-Buslink-FDD1.

Barry Cole
Camas, WA

#6102 - June 2010
AC Motor Control

I’m building an alt-az antenna rotor controller from scratch, but I’m having trouble with the circuit for controlling the AC motor in the RadioShack rotator. I’ve tried using a TRIAC with little success. I believe they are my answer but I don’t understand how to use them. What I need is a circuit that will accept one direction bit and one enable bit that will control a motor’s direction of spin. The original controller achieved this by feeding one or the other non-neutral leads of the motor with ~22V AC; the third lead is common neutral.

David Sarraf
Elizabethtown, PA

#1 You may be having problems with ground referencing the gate signal or with the TRIAC eating up too much of the supply voltage (they do have a small voltage drop when they are ON). One option that would fix both problems would be a pair of mechanical relays. They would exactly duplicate the pushbuttons in the rotor controller, so the rotor will be none the wiser and should work exactly as it always has. You would control the relay coils with a microprocessor and possibly a driver transistor.

Don’t forget to put diodes across the coils so the inductive kick doesn’t hurt the driver or microprocessor output when the relay de-energizes.

If the problem is simply isolation or ground referencing, consider using opto-coupled TRIACS. They work just like the mechanical relay. You put their terminals across the wires that went to the pushbuttons. Their internal LED is the gate drive, so ground referencing is not a factor. They can also be driven from microprocessor outputs. Use a series resistor to limit the current to 20 mA or so. Either solution would be cheap — buy the parts for both and see what works best.

Daniel De Jager
Edgewood, WA

#6103 - June 2010
Quartz Cookoo Clock

My quartz cookoo clock has a photodiode to turn off the clock at night, but I want it to work at night. Can I just remove the photodiode?

First of all, I suspect that the light sensor that turns off the cookoo sound at night is actually a CdS (cadium sulfide) light sensor rather

![Figure 2](image-url)
than a silicon photodiode as you stated. CdS sensors are generally used in consumer products since they are cost-effective and their relatively slow response time (compared to silicon) is acceptable. CdS sensors are light-dependent resistors and can be identified by a serpentine zig-zag sensor pattern. They have a high resistance (in the megohm region) in the dark and a much lower resistance (less than 10K) when illuminated. They are available in various sizes to provide a selection of sensitivity and dynamic range.

You can prevent your clock from going silent at night by simply bridging a resistor (try a value between 1K and 10K) across the CdS sensor in your clock. You could even put a switch in series with the resistor so that the clock could operate either normally or in a modified nighttime mode.

Bob Kovacs
Barnegat, NJ

[#6105 - June 2010]
Vacuum Cleaner Soft Start

I have a Hoover EmPower vacuum cleaner that causes the breakers in my home to trip almost every time I first power it on. After a few minutes of use, I can turn the machine off and back on with no problem. I am thinking that the initial inrush current is too much for the breakers (they are GFCI protected). Does anyone know of some sort of soft start circuit or any other trick that may help?

Len S.
Los Altos, CA

[#2] I doubt it’s the inrush current, but a faulty GFCI. Older GFCIs were prone to tripping with motor loads and/or motors with brushes. I’d suggest replacing the GFCI breaker with a newer one. The specs do list 1,440 watts which does translate to 12A, so I would be attentive to what else is on the circuit.

A metal oxide resistor combined with a delay on make timer such as those made by www.ssac.com would work as a current limiter. The delay on make relay short the resistor out after a specified time delay. Unfortunately, the circuit would have to be at the vacuum cleaner. If the circuit worked, I’d almost be tempted to use a remote AC switch and place the device at the outlet. One example is the remote controls used for woodshop vacuums. I’ve used the technique before and had to experiment with resistor values and time.

Ron Dozier
Wilmington, DE

[#3] There are at least two problems that can cause your circuit breakers to trip when using the vacuum. The first may be a worn out circuit breaker. As circuit breakers age, their springs weaken which causes them to trip on less than their rated current. Frequent tripping can accelerate the aging process. If that’s the case, all you need to do is replace them with new breakers.

Another possibility is a dusty motor. If you have GFCIs, they will trip on leakage current. Dust is hygroscopic which means it will absorb water vapor from the air. That slightly moist dust will leak current which can trip your GFCI. After the motor has run a bit, it warms up and drives off the moisture. The leakage current goes down and the GFCI won’t trip. You didn’t state what was tripping - the overcurrent part or the GFCI part. Either could be the culprit, and the fix should be fairly easy. I do not believe it is from too much current draw. The vacuum is rated for 12 amps which means that it should be okay on a standard 15 or 20 amp circuit. Motors do draw more on startup, but consider that this is a widely-used consumer product and a Google search shows nobody else having a problem with it.

David Sarraf
Elizabethtown, PA

[#4] Ground Fault Interrupt (GFI) breakers are only installed in kitchen, bath, and outdoor outlets. Assuming that you are plugged into a bedroom or living room outlet, the GFI breaker is not the source of your problem. Do you have a relatively new house with arc fault breakers? The arc fault detecting breaker is known to sometimes trip on brushed motors such as used in vacuum cleaners. Try the vacuum cleaner on the outlet serving your refrigerator. It will not have an arc fault breaker because the reliable operation of the refrigerator trumps safety considerations. If the vacuum cleaner does not trip the refrigerator outlet, it is not the starting surge current; you need to plug into something other than an arc fault breaker.

Dennis Crunkilton
via email

[#6104 - June 2010]
Driveway Sensor

I want to set up a home driveway sensor similar to a traffic light actuator by using a single wire loop in the pavement that senses vehicles by changing the frequency of an oscillator. I need to know how it works and how to build it.

The unit works very much like a large metal detector, although the traffic signal units have more features.

I have access to many of these units if you have an interest. The traffic units have the ability to pulse (pulse the relay if a vehicle passes over or sits on the loop) a built-in relay, or they can place a constant signal to the relay (if a vehicle is sitting on the loop). Some have the ability to delay or extend the detected vehicle. Your loop will have to be about 4-6 turns of 14 Ga. wire in a 6" x 6" or smaller area. The detector also has frequency and sensitivity, and will let you know of an out of range or faulty loop.

Ray
via email
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## October 2010 NUTS & VOLTS 81
Triple Output Bench Power Supplies

The new CSI3303S & CSI5505S regulated DC power supplies are high reliability, variable DC Power Supplies with built-in short circuit and thermal protection. These power supplies are suitable for the laboratory, electronics, communications equipment maintenance, production line, scientific research and educational institutions. Both units are equipped with protection circuits that protect the units from short circuits and over temperature by shutting the unit down for safety. Both units allow independent, serial and parallel mode operation.

Technical Specifications:

- Independent mode: 2 independent 0-30V outputs
- Series mode: CSI3303S Output from 0-60V & 0-3A
- CSI5505S Output from 0-60V & 0-5A
- Parallel mode: CSI3303S Output from 0-6A & 0-30V
- CSI5505S Output from 0-10A & 0-30V
- Both units also provide a 5V fixed output @ 3A
- Load regulation: <0.1%+3mA (rating current <3A)
- 2%+3mA
- Ripple and noise: <1mVRms 5Hz-1MHz
- <3mArms
- Voltage accuracy: +/-0.5%rdg+2byte
- Current accuracy: +/-5%rdg+2byte
- Display resolution: +/-0.5%rdg+2byte
- Rated output: 5.0V +/-0.1V 3A
- Tracking characteristics:
- Series specifications:
  - Load regulation: less than 50mV
  - Ripple and noise: (5Hz-1MHz) <=3mVRMS
- Parallel characteristics:
  - Load regulation: less than 50mV
  - Ripple and noise: (5Hz-1MHz) CV less than
  - 1mV=6A, CV less than
  - 1.5mV (p-6A)

Once again Circuit Specialists brings you a quality product at a great price!

USB Digital Storage Oscilloscopes

Specifications

40MHz
- DSO-2090
- DSO-2150
- DSO-2250
- DSO-5200
- DSO-5200A

60MHz
- DSO-2090
- DSO-2150
- DSO-2250
- DSO-5200
- DSO-5200A

100MHz
- DSO-2090
- DSO-2150
- DSO-2250
- DSO-5200
- DSO-5200A

200MHz
- DSO-2090
- DSO-2150
- DSO-2250
- DSO-5200
- DSO-5200A

355MHz
- DSO-2090
- DSO-2150
- DSO-2250
- DSO-5200
- DSO-5200A

Full specifications at www.CircuitSpecs.com/Aardvark

Wireless Inspection Camera

With Color 3.5" LCD Recordable Monitor

Your Extended Eyes and Hands!

Records

Still Pictures & Video

See It!
Clearly in narrow spots, even in total darkness or underwater.

Find It!
Fast. No more struggling with a mirror & flashlight.

Solve It!
Easily, speed up the solution with extended accessories.

Record It!
With the 3.5" LCD recordable monitor, you can capture pictures or record video for documentation.

Full specifications at www.CircuitSpecs.com/Aardvark

Item #
AARDVARK

3ft Extension AARDVARK-EXT $24.95

Probes Included.
We carry a LARGE selection of Power Supplies, Soldering Equipment, Test Equipment, Oscilloscopes, Digital Multimeters, Electronic Components, Metal and Plastic Project Boxes, Electronic Chemicals, PC Based Digital I/O Cards, Panel Meters, Breadboards, Device Programmers, and many other interesting items. Check out our website at: www.CircuitSpecialists.com

19inch Rack Mount Switch-Mode Power Supply with 3 Output Modes

The BTI-150 is a high efficiency switching power supply, for use where multiple voltage continuous power output is needed. It is ideal for use in Technical Product Research, Laboratory, Educational Institutions, Telecommunication Industry, etc.

The Rack Mount Design makes it easy to stack multiple power supplies for high use areas.

<table>
<thead>
<tr>
<th>Output Voltage</th>
<th>Output Current</th>
<th>V/C Display</th>
<th>Resolution</th>
</tr>
</thead>
<tbody>
<tr>
<td>0-15V</td>
<td>0-10A</td>
<td>Digital</td>
<td>0.1%/±2%</td>
</tr>
<tr>
<td>0-30V</td>
<td>0-5A</td>
<td>Digital</td>
<td>0.1%/±2%</td>
</tr>
<tr>
<td>0-60V</td>
<td>0-2.5A</td>
<td>Digital</td>
<td>0.1%/±2%</td>
</tr>
</tbody>
</table>

Item # BTI-150

$189.00

www.circuitspecialists.com/BI-150

200MHz Hand Held Scopemeter with Oscilloscope & DMM Functions

You get both a 200 MHz Oscilloscope and a multifunction digital multimeter all in one convenient lightweight rechargeable battery powered package. This powerful compact package comes complete with scope meter, test leads, two 6-inch probes, chargers, PC software, USB cable and a convenient nylon carrying case.

DC Voltage: 0-18V
DC Current: 5A
Power (max): 90W

Item # DSO1200

$739.00

www.circuitspecialists.com/DSO1200

60MHz Hand Held Scopemeter with Oscilloscope & DMM Functions

Get this powerful handheld digital multimeter with integrated digital multimeter support. 60MHz Bandwidth with 2 Channels. 0.5% High Definition Sampling Rate. 500mA’s Equivalent-Time Sampling Rate. 6,000-Count DMM resolution with AC/DC at 600V/800V, 10A. Large 5.7 inch TFT Color LCD Display. USB Host/Device 2.0 Full-speed Interface Connectivity. Multi-Language Support. Battery Power Operation (Installed).

Item # DSO1060

$529.00

www.circuitspecialists.com/DSO1060

60MHz Hand Held Scopemeter w/Oscilloscope, DMM Functions & 25 MHz Arbitrary Waveform Generator

The features of the DSO-8060 plus a 25 MHz Arbitrary Waveform Generator.

- Waveforms can be saved in the following formats: .jpg, .bmp, .csv, .txt, .wav, .mp4, .avi
- Can record and save 1000 waveforms
- DC to 25 MHz Arbitrary Waveform Generator

Item # DSO-8060

$659.00

www.circuitspecialists.com/DSO-8060

Programmable DC Electronic Loads

These devices can be used with supplies up to 360VDC and 30A. It features a rotary selection switch and a numeric keypad used to input the maximum voltage, current and power settings. These electronic DC loads are perfect for use in laboratory environments and schools, or for testing DC power supplies or high-capacity batteries. It also features memory, and can also be connected to a PC, to implement remote control and supervision.

360V/150W (CSI3710A) $349.00
www.circuitspecialists.com/CSI3710A

360V/300W (CSI3711A) $499.00
www.circuitspecialists.com/CSI3711A

Programmable DC Power Supplies

- Up to 10 settings stored in memory
- Optional RS-232, USB, RS-485 adapters
- May be used in series or parallel modes with additional supplies.
- Low output ripple & noise
- LCD display with backlight
- High resolution at 1mV

<table>
<thead>
<tr>
<th>Model</th>
<th>CSI3644A</th>
<th>CSI3645A</th>
<th>CSI3646A</th>
</tr>
</thead>
<tbody>
<tr>
<td>DC Voltage</td>
<td>0-18V</td>
<td>0-36V</td>
<td>0-72V</td>
</tr>
<tr>
<td>DC Current</td>
<td>5A</td>
<td>3A</td>
<td>1.5A</td>
</tr>
<tr>
<td>Power (max)</td>
<td>90W</td>
<td>108W</td>
<td>108W</td>
</tr>
<tr>
<td>Price</td>
<td>$199.00</td>
<td>$199.00</td>
<td>$199.00</td>
</tr>
</tbody>
</table>

Item # CS1530S

$84.95

www.circuitspecialists.com/CSI530S
The Propeller-powered AP-16+ (#31.316; $129.99) by EFX-TEK allows you to play back high-quality, stereo WAV files for your standard SD card through two, booming 20 W amplifiers (one per channel). Advanced features like built-in randomization, auxiliary inputs, and the ability to use dry-contact, (optically isolated) 12 to 24 DC, or a PIR (or compatible) sensor put the AP-16+ in a class of its own. For standalone control the AP-16+ includes pre- and post-play delay potentiometers, as well as a relay that is active while your audio plays. Need background audio while your animated exhibit is idle? No problem, the AP-16+ includes an ambient loop control to do just that.

Features:
- 8-core Propeller P8X32A-Q44 controller on board
- Plays stereo WAV files (16-bit, PCM, up to 44.1 kHz)
- Files stored on your standard SD card (FAT or FAT32 format)
- 2.1 mm, center-positive connector for DC power
- 24 stand-alone files under manual control
- Optional, auto-playing ambient file (AMBIENT.WAV)
- Manual input via (normally-open) dry-contact, (optically isolated) 12-24 VDC input, or PIR sensor
- Eight, dry-contact auxiliary inputs
- Relay output active when file plays (N.O. and N.C connections provided)
- Pre- and post-play delay potentiometers (manual mode)
- Dual 20 W amplifiers (one per channel)
- TTL serial control for connection to host microcontroller — unlimited files under serial control
- User-configured baud rate for serial control
- User-configured address allows up to four AP-16+ boards on one serial connection
- Easily updated as new features become available — update loads from an SD card (not included)
- Four 0.135 inch (3.4 mm) mounting holes

An 18 V 3.3 Amp Power Supply (sold separately; #750-00011; $34.99) is recommended for use with this unit.

PIR Sensor (#555-28027; $9.99) sold separately.

Order the Propeller-powered AP-16+ Audio Player online at www.parallax.com or call toll-free: 888-512-1024 (M-F, 7am-5pm, PDT).